



7MHz, Rail-to-Rail I/O CMOS Operational Amplifier

1 FEATURES

- HIGH GAIN BANDWIDTH:7MHz
- RAIL-TO-RAIL INPUT AND OUTPUT ±0.1mV Typical Vos
- INPUT VOLTAGE RANGE: -0.1V to +5.6V with Vs = 5.5V
- SUPPLY RANGE: +2.5V to +5.5V
- SPECIFIED UP TO +125°C
- Micro SIZE PACKAGES: SOT23-5

2 APPLICATIONS

- SENSORS
- PHOTODIODE AMPLIFICATION
- ACTIVE FILTERS
- TEST EQUIPMENT
- DRIVING A/D CONVERTERS

3 DESCRIPTIONS

The RS62XP families of products offer low voltage operation and rail-to-rail input and output, as well as excellent speed/power consumption ratio, providing an excellent bandwidth (7MHz) and slew rate of 4.3V/us. The op-amps are unity gain stable and feature an ultra-low input bias current.

The devices are ideal for sensor interfaces, active filters and portable applications. The RS62XP families of operational amplifiers are specified at the full temperature range of -40°C to +125°C under single or dual power supplies of 2.5V to 5.5V.

PART NUMBER	PACKAGE	BODY SIZE(NOM)	
	SOT23-5	2.90mm×1.60mm	
DC / 01 D	SOP8	4.90mm×3.90mm	
RS621P	MSOP8	3.00mm×3.00mm	
	SC70-5	2.10mm×1.25mm	
	SOP8	4.90mm×3.90mm	
RS622P	MSOP8	3.00mm×3.00mm	
K2022P	TSSOP8	3.00mm×4.40mm	
	DFN2X2-8	2.00mm×2.00mm	
RS624P	SOP14	8.65mm×3.90mm	
	TSSOP14	5.00mm×4.40mm	

Device Information⁽¹⁾

(1) For all available packages, see the orderable addendum at the end of the data sheet.



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4 Revision History Note: Page numbers for previous revisions may different from page numbers in the current version.

Version	Change Date	Change Item
D.1	2022/07/19	Version Updated
D.2	2023/09/20	Added Thermal Pad Pin Description
D.2.1	2024/03/01	Modify packaging naming



5 PACKAGE/ORDERING INFORMATION⁽¹⁾

Orderable Device	Package Type	Pin	Channel	Op Temp(°C)	Device Marking ⁽²⁾	Package Qty
RS621PXF	SOT23-5	5	1	-40°C ~125°C	621P	Tape and Reel,3000
RS621PXC5	SC70-5 ⁽³⁾	5	1	-40°C ~125°C	621P	Tape and Reel,3000
RS621BPXF	SOT23-5	5	1	-40°C ~125°C	621BP	Tape and Reel,3000
RS621BPXC5	SC70-5 ⁽³⁾	5	1	-40°C ~125°C	621BP	Tape and Reel,3000
RS621PXK	SOP8	8	1	-40°C ~125°C	RS621P	Tape and Reel,4000
RS621PXM	MSOP8	8	1	-40°C ~125°C	RS621P	Tape and Reel,4000
RS622PXK	SOP8	8	2	-40°C ~125°C	RS622P	Tape and Reel,4000
RS622PXM	MSOP8	8	2	-40°C ~125°C	RS622P	Tape and Reel,4000
RS622PXQ	TSSOP8	8	2	-40°C ~125°C	RS622P	Tape and Reel,4000
RS622PXTDE8	DFN2X2-8	8	2	-40°C ~125°C	622P	Tape and Reel,3000
RS624PXP	SOP14	14	4	-40°C ~125°C	RS624P	Tape and Reel,4000
RS624PXQ	TSSOP14	14	4	-40°C ~125°C	RS624P	Tape and Reel,4000

NOTE:

(1) This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the right-hand navigation.

(2) There may be additional marking, which relates to the lot trace code information (data code and vendor code), the logo or the environmental category on the device.

(3) Equivalent to SOT353.



6 Pin Configuration and Functions (Top View)



Pin Description

		PIN			
NAME	RS621P	RS621BP	RS621P	I/O ⁽¹⁾	DESCRIPTION
	SOT23-5/SC70-5	SOT23-5/SC70-5	SOP8/MSOP8		
-IN	4	3	2	I	Negative (inverting) input
+IN	3	1	3	I	Positive (noninverting) input
NC ⁽²⁾	-	-	1,5,8	-	No internal connection (can be left floating)
OUT	1	4	6	0	Output
V-	2	2	4	-	Negative (lowest) power supply
V+	5	5	7	-	Positive (highest) power supply

(1) I = Input, O = Output.

(2) There is no internal connection. Typically, GND is the recommended connection to a heat spreading plane.





Pin Description

	PIN SOP8/MSOP8/TSSOP8/DFN2X2-8	I/O ⁽¹⁾	DESCRIPTION
-INA	2	I	Inverting input, channel A
+INA	3	I	Noninverting input, channel A
-INB	6	I	Inverting input, channel B
+INB	5	I	Noninverting input, channel B
OUTA	1	0	Output, channel A
OUTB	7	0	Output, channel B
V-	4	-	Negative (lowest) power supply
V+	8	-	Positive (highest) power supply
-	Thermal Pad	-	Connect thermal pad to V-
1) I = Input	t, O = Output.	•	•



Pin Configuration and Functions (Top View)



Pin Description

	PIN		DESCRIPTION	
NAME	SOP14/TSSOP14	I/O ⁽¹⁾	DESCRIPTION	
-INA	2	I	Inverting input, channel A	
+INA	3	I	Noninverting input, channel A	
-INB	6	I	Inverting input, channel B	
+INB	5	I	Noninverting input, channel B	
-INC	9	I	Inverting input, channel C	
+INC	10	I	Noninverting input, channel C	
-IND	13	I	Inverting input, channel D	
+IND	12	I	Noninverting input, channel D	
OUTA	1	0	Output, channel A	
OUTB	7	0	Output, channel B	
OUTC	8	0	Output, channel C	
OUTD	14	0	Output, channel D	
V-	11	-	Negative (lowest) power supply	
V+	4	-	Positive (highest) power supply	

(1) I = Input, O = Output,



7 SPECIFICATIONS

7.1 Absolute Maximum Ratings

Over operating free-air temperature range (unless otherwise noted)⁽¹⁾

			MIN	MAX	UNIT
	Supply, V _S =(V+) - (V-)		7		
Voltage	Signal input pin ⁽²⁾		(V-)-0.5	(V+) +0.5	V
	Signal output pin ⁽³⁾		(V-)-0.5	(V+) +0.5	
	Signal input pin ⁽²⁾		-10	10	mA
Current	Signal output pin ⁽³⁾		-150	150	mA
	Output short-circuits (4)	Cont	inuous		
		SOT23-5		230	
	Package thermal impedance ⁽⁵⁾	SOP8		110.88	°C/W
		MSOP8		165.7	
0		TSSOP8		240	
ALθ		SOP14		104.5	°C/W
		TSSOP14		89.21	
		SC70-5		376	
		DFN2X2-8		80	
	Operating range, T _A		-40	125	
Temperature	Junction, T ⁽⁶⁾		-40	150	°C
	Storage, T _{stg}	-65	150		

(1) Stresses above these ratings may cause permanent damage. Exposure to absolute maximum conditions for extended periods may degrade device reliability. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those specified is not implied.

(2) Input terminals are diode-clamped to the power-supply rails. Input signals that can swing more than 0.5V beyond the supply rails should be current-limited to 10mA or less.

(3) Output terminals are diode-clamped to the power-supply rails. Output signals that can swing more than 0.5V beyond the supply rails should be current-limited to ±150mA or less.

(4) Short-circuit to ground, one amplifier per package.

(5) The package thermal impedance is calculated in accordance with JESD-51.

(6) The maximum power dissipation is a function of $T_{J(MAX)}$, $R_{\theta JA}$, and T_A . The maximum allowable power dissipation at any ambient temperature is $P_D = (T_{J(MAX)} - T_A) / R_{\theta JA}$. All numbers apply for packages soldered directly onto a PCB.

7.2 ESD Ratings

The following ESD information is provided for handling of ESD-sensitive devices in an ESD protected area only.

			VALUE	UNIT
		Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 ⁽¹⁾	±5000	
V(ESD)	Electrostatic discharge	Charged-device model (CDM), per ANSI/ESDA/JEDEC JS-002 $^{(2)}$	±1000	V
		Machine Model (MM)	±400	

(1) JEDEC document JEP155 states that 500 V HBM allows safe manufacturing with a standard ESD control process.
 (2) JEDEC document JEP157 states that 250 V CDM allows safe manufacturing with a standard ESD control process.



ESD SENSITIVITY CAUTION

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.



7.3 Recommended Operating Conditions Over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
Supply voltage , V _S = (V+) - (V-)	Single-supply	2.5		5.5	V
	Dual-supply	±1.25		±2.75	v



7.4 ELECTRICAL CHARACTERISTICS

(At $T_A = +25$ °C, Vs=2.5V to 5V, R_L = 10k Ω connected to V_S/2, and V_{OUT} = V_S/2, V_{CM} = V_S/2, Full ⁽⁹⁾ = -40°C to +125°C, unless otherwise noted.) ⁽¹⁾

	DADAMETED	CONDITIONS	т.	RS621P, RS622P, RS624P			
	PARAMETER	CONDITIONS	T,	MIN ⁽²⁾	TYP ⁽³⁾	MAX ⁽²⁾	UNIT
POWER	SUPPLY						
Vs	Operating Voltage Range		25°C	2.5		5.5	V
lq	Quiescent Current Per Amplifier	Vs=±2.5V, Io=0mA	25°C		625	850	uA
	Dower Surphy Dejection Datio	$\lambda = 2.5 \lambda + 2.5 \lambda$	25°C	75	90		٩Þ
PSRR	Power-Supply Rejection Ratio	V _s =2.5V to 5.5V	Full	68			dB
INPUT							
Vos	Input Offset Voltage	$V_{CM}=V_S/2$	25°C	-0.8	±0.1	0.8	mV
Vos Tc	Input Offset Voltage Average Drift	V _{CM} =V _S /2	Full		±2.3		uV/°C
IB	Input Bias Current ^{(4) (5)}	V _{CM} =0V	25°C		±1	±10	pА
los ⁽⁵⁾	Input Offset Current	V _{CM} =0V	25°C		±1	±10	pА
V_{CM}	Common-Mode Voltage Range	V _S = 5.5V	25°C	-0.1		5.6	V
	Common-Mode Rejection Ratio	Vs = 5.5V V _{CM} =-0.1V to 3.5V	25°C	73	94		dB
CMRR			Full	64			
		V _S = 5.5V V _{CM} =-0.1V to 5.6V	25°C	60	80		
			Full	59			
OUTPU	Т						
Aol		R _L =10KΩ, V _O =(V-)	25°C	110	127		dB
AOL	Open-Loop Voltage Gain	+0.1V to (V+)-0.1V	Full	98			uБ
	Output Swing From Rail	$V_S=\pm 2.5V, R_L=10K\Omega$	25°C		10	20	mV
Ιουτ	Output Short-Circuit Current ^{(6) (7)}		25°C	±50	±102		mA
C_{LOAD}	Capacitive Load Drive				100		pF
FREQUE	ENCY RESPONSE						
SR	Slew Rate ⁽⁸⁾	G=+1, CL=100pF	25°C		4.3		V/us
GBP	Gain-Bandwidth Product		25°C		7		MHz
PM	Phase Margin ⁽⁵⁾		25°C		64		0
ts	Settling Time,0.1%	$V_S=\pm 2.5V, G=+1,$ $C_L=100pF,$ Step=2V	25°C		1.9		us
tor	Overload Recovery Time	V _{IN} •Gain≥V _S , G=-10	25°C		0.45		us
NOISE	•	·		•		•	
En	Input Voltage Noise	f = 0.1Hz to 10Hz, Vs=±2.5V	25°C		3.6		uV _{PP}
en	Input Voltage Noise Density	f = 1KHz	25°C		TBD		nV/√Hz

(1) Electrical table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device.

(2) Limits are 100% production tested at 25°C. Limits over the operating temperature range are ensured through correlations using statistical quality control (SQC) method.

(3) Typical values represent the most likely parametric norm as determined at the time of characterization. Actual typical values may vary over time and will also depend on the application and configuration.

(4) Positive current corresponds to current flowing into the device.

(5) This parameter is ensured by design and/or characterization and is not tested in production.

(6) The maximum power dissipation is a function of $T_{J(MAX)}$, R_{0JA} , and T_A . The maximum allowable power dissipation at any ambient temperature is PD = ($T_{J(MAX)} - T_A$) / R_{0JA} . All numbers apply for packages soldered directly onto a PCB.

(7) Short circuit test is a momentary test.

(8) Number specified is the slower of positive and negative slew rates.

(9) Specified by characterization only.



7.5 TYPICAL CHARACTERISTICS

NOTE: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only.

At $T_A = +25^{\circ}$ C, $V_S=5V$, $R_L = 10k\Omega$ connected to $V_S/2$, $V_{OUT} = V_S/2$, unless otherwise noted.







Figure 3. Quiescent Current vs Temperature



Figure 5. Small-Signal Step Response



Figure 2. Power-Supply Rejection Ratio vs Frequency



Figure 4. Input Bias Current vs Temperature



Figure 6. Large-Signal Step Response



TYPICAL CHARACTERISTICS

NOTE: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only.

At $T_A = +25^{\circ}$ C, $V_S=5V$, $R_L = 10k\Omega$ connected to $V_S/2$, $V_{OUT} = V_S/2$, unless otherwise noted.



Figure 7. Negative Overvoltage Recovery



Figure 9. 0.1Hz to 10Hz Input Voltage Noise



Figure 8. Positive Overvoltage Recovery



8 Detailed Description

8.1 Overview

The RS62XP are high precision, rail-to-rail operational amplifiers that can be run from a single-supply voltage 2.5V to 5.5V (\pm 1.25V to \pm 2.75V). Supply voltages higher than 7V (absolute maximum) can permanently damage the amplifier. Rail-to-rail input and output swing significantly increases dynamic range, especially in low-supply applications. Good layout practice mandates use of a 0.1uF capacitor place closely across the supply pins.

8.2 Phase Reversal Protection

The RS62XP family has internal phase-reversal protection. Many op amps exhibit phase reversal when the input is driven beyond the linear common-mode range. This condition is most often encountered in noninverting circuits when the input is driven beyond the specified common-mode voltage range, causing the output to reverse into the opposite rail. The input of the RS62XP prevents phase reversal with excessive common-mode voltage. Instead, the appropriate rail limits the output voltage. This performance is shown in figure 10.



Figure 10. Output Waveform Devoid of Phase Reversal During an Input Overdrive Condition

8.3 EMI Rejection Ratio (EMIRR)

The electromagnetic interference (EMI) rejection ratio, or EMIRR, describes the EMI immunity of operational amplifiers. An adverse effect that is common to many operational amplifiers is a change in the offset voltage as a result of RF signal rectification. An operational amplifier that is more efficient at rejecting this change in offset as a result of EMI has a higher EMIRR and is quantified by a decibel value. Measuring EMIRR can be performed in many ways, but this document provides the EMIRR IN+, which specifically describes the EMIRR performance when the RF signal is applied to the noninverting input pin of the operational amplifier. In general, only the noninverting input is tested for EMIRR for the following three reasons:

• Operational amplifier input pins are known to be the most sensitive to EMI, and typically rectify RF signals better than the supply or output pins.

• The noninverting and inverting operational amplifier inputs have symmetrical physical layouts and exhibit nearly matching EMIRR performance.

• EMIRR is easier to measure on noninverting pins than on other pins because the noninverting input pin can be isolated on a printed-circuit-board (PCB). This isolation allows the RF signal to be applied directly to the noninverting input pin with no complex interactions from other components or connecting PCB traces.

The EMIRR IN+ of the RS62XP is plotted versus frequency in Figure 11. If available, any dual and quad operational amplifier device versions have approximately identical EMIRR IN+ performance. The RS62XP unity-gain bandwidth is 7MHz. EMIRR performance below this frequency denotes interfering signals that fall within the operational amplifier bandwidth.



Detailed Description (continued)



Figure 11. RS62XP EMIRR vs Frequency

8.4 EMIRR IN+ Test Configuration

Figure 12 shows the circuit configuration for testing the EMIRR IN+. An RF source is connected to the operational amplifier noninverting input pin using a transmission line. The operational amplifier is configured in a unity-gain buffer topology with the output connected to a low-pass filter (LPF) and a digital multimeter (DMM). A large impedance mismatch at the operational amplifier input causes a voltage reflection; however, this effect is characterized and accounted for when determining the EMIRR IN+. The resulting dc offset voltage is sampled and measured by the multimeter. The LPF isolates the multimeter from residual RF signals that can interfere with multimeter accuracy.



Figure 12. EMIRR IN+ Test Configuration Schematic



9 Application and Implementation

Information in the following applications sections is not part of the RUNIC component specification, and RUNIC does not warrant its accuracy or completeness. RUNIC's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

9.1 APPLICATION NOTE

The RS62XP are high precision, rail-to-rail operational amplifiers that can be run from a single-supply voltage 2.5V to 5.5V (\pm 1.25V to \pm 2.75V). Supply voltages higher than 7V (absolute maximum) can permanently damage the amplifier. Rail-to-rail input and output swing significantly increases dynamic range, especially in low-supply applications. Good layout practice mandates use of a 0.1uF capacitor place closely across the supply pins.

Typical Applications 9.2 25-kHz Low-pass Filter



Figure 13. 25-kHz Low-Pass Filter

9.3 Design Requirements

Low-pass filters are commonly employed in signal processing applications to reduce noise and prevent aliasing. The RS62XP devices are ideally suited to construct high-speed, high-precision active filters. Figure 13 shows a second-order, low-pass filter commonly encountered in signal processing applications.

Use the following parameters for this design example:

- Gain = 5 V/V (inverting gain)
- Low-pass cutoff frequency = 25 kHz
- Second-order Chebyshev filter response with 3-dB gain peaking in the passband

9.4 Detailed Design Procedure

The infinite-gain multiple-feedback circuit for a low-pass network function is shown in Figure 13. Use Equation 1 to calculate the voltage transfer function.

$$\frac{\text{Output}}{\text{Input}}(s) = \frac{-1/R_1R_3C_2C_5}{s^2 + (s/C_2)(1/R_1 + 1/R_3 + 1/R_4) + 1/R_3R_4C_2C_5}$$

This circuit produces a signal inversion. For this circuit, the gain at dc and the low-pass cutoff frequency are calculated by Equation 2:

$$Gain = \frac{R_4}{R_1}$$
$$f_C = \frac{1}{2\pi} \sqrt{(1/R_3R_4C_2C_5)}$$

(2)

(1)



9.5 Application Curve



Figure 14. Low-pass filter transfer function



10 LAYOUTS

10.1 Layout Guidelines

Attention to good layout practices is always recommended. Keep traces short. When possible, use a PCB ground plane with surface-mount components placed as close to the device pins as possible. Place a 0.1uF capacitor closely across the supply pins.

These guidelines should be applied throughout the analog circuit to improve performance and provide benefits such as reducing the EMI susceptibility.

10.2 Layout Example



Figure 15. Schematic Representation





NOTE: Layout Recommendations have been shown for dual op-amp only, follow similar precautions for Single and four.



11 PACKAGE OUTLINE DIMENSIONS SOT23-5⁽³⁾





RECOMMENDED LAND PATTERN (Unit: mm)





Symphol	Dimensions I	n Millimeters	Dimensions In Inches		
Symbol	Min	Max	Min	Мах	
A ⁽¹⁾	1.050	1.250	0.041	0.049	
A1	0.000	0.100	0.000	0.004	
A2	1.050	1.150	0.041	0.045	
b	0.300	0.500	0.012	0.020	
с	0.100	0.200	0.004	0.008	
D ⁽¹⁾	2.820	3.020	0.111	0.119	
E ⁽¹⁾	1.500	1.700	0.059	0.067	
E1	2.650	2.950	0.104	0.116	
e	0.950(BSC) ⁽²⁾	0.037(BSC) ⁽²⁾	
e1	1.800	2.000	0.071	0.079	
L	0.300	0.600	0.012	0.024	
θ	0°	8°	0°	8°	

NOTE:

1. Plastic or metal protrusions of 0.15mm maximum per side are not included.

2. BSC (Basic Spacing between Centers), "Basic" spacing is nominal.
 3. This drawing is subject to change without notice.



SC70-5⁽³⁾





RECOMMENDED LAND PATTERN (Unit: mm)





Symbol	Dimensions I	n Millimeters	Dimension	s In Inches
Symbol	Min	Max	Min 0.035 0.000 0.035 0.006 0.003 0.003 0.0045 0.085 0.026(Мах
A (1)	0.900	1.100	0.035	0.043
A1	0.000	0.100	0.000	0.004
A2	0.900	1.000	0.035	0.039
b	0.150	0.350	0.006	0.014
с	0.080	0.150	0.003	0.006
D ⁽¹⁾	2.000	2.200	0.079	0.087
E ⁽¹⁾	1.150	1.350	0.045	0.053
E1	2.150	2.450	0.085	0.096
е	0.650(BSC) ⁽²⁾	0.026(BSC) ⁽²⁾
e1	1.300(1.300(BSC) (2)		BSC) ⁽²⁾
L	0.260	0.460	0.010	0.018
L1	0.5	525	0.0	021
θ	0°	8°	0°	8°

NOTE:

1. Plastic or metal protrusions of 0.15mm maximum per side are not included.

2. BSC (Basic Spacing between Centers), "Basic" spacing is nominal.

3. This drawing is subject to change without notice.



MSOP8⁽³⁾





RECOMMENDED LAND PATTERN (Unit: mm)





Symbol	Dimensions I	n Millimeters	Dimension	s In Inches
Symbol	Min	Max	Min	Мах
A ⁽¹⁾	0.820	1.100	0.032	0.043
A1	0.020	0.150	0.001	0.006
A2	0.750	0.950	0.030	0.037
b	0.250	0.380	0.010	0.015
С	0.090	0.230	0.004	0.009
D ⁽¹⁾	2.900	3.100	0.114	0.122
е	0.650(BSC) ⁽²⁾	0.026(BSC) ⁽²⁾
E ⁽¹⁾	2.900	3.100	0.114	0.122
E1	4.750	5.050	0.187	0.199
L	0.400	0.800	0.016	0.031
θ	0°	6°	0°	6°

NOTE:

1. Plastic or metal protrusions of 0.15mm maximum per side are not included.

2. BSC (Basic Spacing between Centers), "Basic" spacing is nominal.

3. This drawing is subject to change without notice.

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TSSOP14⁽³⁾





RECOMMENDED LAND PATTERN (Unit: mm)





Sumahal	Dimensions I	n Millimeters	Dimension	s In Inches	
Symbol	Min	Max	Min	Мах	
A ⁽¹⁾		1.200		0.047	
A1	0.050	0.150	0.002	0.006	
A2	0.800	1.050	0.031	0.041	
b	0.190	0.300	0.007	0.012	
с	0.090	0.200	0.004	0.008	
D ⁽¹⁾	4.860	5.100	0.191	0.201	
E ⁽¹⁾	4.300	4.500	0.169	0.177	
E1	6.250	6.550	0.246	0.258	
e	0.650(BSC) ⁽²⁾	0.026(BSC) ⁽²⁾	
L	0.500	0.700	0.020	0.028	
Н	0.25	(TYP)	0.01	(TYP)	
θ	1°	7°	1°	7°	

NOTE:

. 1. Plastic or metal protrusions of 0.15mm maximum per side are not included. 2. BSC (Basic Spacing between Centers), "Basic" spacing is nominal. 3. This drawing is subject to change without notice.



SOP8⁽³⁾





RECOMMENDED LAND PATTERN (Unit: mm)





Symbol	Dimensions I	n Millimeters	Dimensions In Inches			
Symbol	Min	Max	Dimensions In In Min 0.053 0.004 0.053 0.013 0.007 0.189 0.0228 0.150	Max		
A ⁽¹⁾	1.350	1.750	0.053	0.069		
A1	0.100	0.250	0.004	0.010		
A2	1.350	1.550	0.053	0.061		
b	0.330	0.510	0.013	0.020		
с	0.170	0.250	0.007	0.010		
D ⁽¹⁾	4.800	5.000	0.189	0.197		
e	1.270(BSC) ⁽²⁾	0.050(BSC) ⁽²⁾		
E	5.800	6.200	0.228	0.244		
E1 ⁽¹⁾	3.800	4.000	0.150	0.157		
L	0.400	1.270	0.016	0.050		
θ	0°	8°	0°	8°		

NOTE:

1. Plastic or metal protrusions of 0.15mm maximum per side are not included.
2. BSC (Basic Spacing between Centers), "Basic" spacing is nominal.
3. This drawing is subject to change without notice.



SOP14⁽³⁾





RECOMMENDED LAND PATTERN (Unit: mm)





Sympol	Dimensions I	n Millimeters	Dimension	s In Inches
Symbol	Min	Max	Dimensions I Min 0.053 0.004 0.053 0.012 0.004 0.004 0.012 0.004 0.004 0.0053 0.004 0.005 0.004 0.0333 0.228 0.150	Мах
A ⁽¹⁾	1.350	1.750	0.053	0.069
A1	0.100	0.250	0.004	0.010
A2	1.350	1.550	0.053	0.061
b	0.310	0.510	0.012	0.020
с	0.100	0.250	0.004	0.010
D ⁽¹⁾	8.450	8.850	0.333	0.348
e	1.270(BSC) ⁽²⁾	0.050(BSC) ⁽²⁾
E	5.800	6.200	0.228	0.244
E1 ⁽¹⁾	3.800	4.000	0.150	0.157
L	0.400	1.270	0.016	0.050
θ	0°	8°	0°	8°

NOTE:

1. Plastic or metal protrusions of 0.15mm maximum per side are not included.

2. BSC (Basic Spacing between Centers), "Basic" spacing is nominal.

3. This drawing is subject to change without notice.



TSSOP8⁽³⁾





RECOMMENDED LAND PATTERN (Unit: mm)





Cumhal	Dimensions I	n Millimeters	Dimension	s In Inches
Symbol	Min	Max	Min	Max
A ⁽¹⁾		1.200		0.047
A1	0.050	0.150	0.002	0.006
A2	0.800	1.050	0.031	0.041
b	0.190	0.300	0.007	0.012
с	0.090	0.200	0.004	0.008
D ⁽¹⁾	2.900	3.100	0.114	0.122
E ⁽¹⁾	4.300	4.500	0.169	0.177
E1	6.250	6.550	0.246	0.258
e	0.650(BSC) ⁽²⁾	0.026(BSC) ⁽²⁾
L	0.500	0.700	0.020	0.028
Н	0.25	(TYP)	0.01	(TYP)
θ	1°	7°	1°	7°

NOTE:

1. Plastic or metal protrusions of 0.15mm maximum per side are not included.

2. BSC (Basic Spacing between Centers), "Basic" spacing is nominal.
 3. This drawing is subject to change without notice.



DFN2X2-8⁽²⁾



RECOMMENDED LAND PATTERN (Unit: mm)

Sumbol	Dimensions I	n Millimeters	Dimensions In Inches			
Symbol	Min	Max	Min N 0.028 0. 0.000 0. 0.000 0. 0.007 0. 0.075 0. 0.075 0. 0.075 0. 0.075 0. 0.024 0.	Мах		
A ⁽¹⁾	0.700	0.800	0.028	0.031		
A1	0.000	0.050	0.000	0.002		
A2	0.203	(TYP)	0.008(TYP)			
b	0.180	0.300	0.007	0.012		
D ⁽¹⁾	1.900	2.100	0.075	0.083		
D1	1.100	1.300	0.043	0.051		
E ⁽¹⁾	1.900	2.100	0.075	0.083		
E1	0.600	0.800	0.024	0.031		
e	0.500)(TYP)	0.020	(TYP)		
L	0.250	0.450	0.010	0.018		

NOTE:

Plastic or metal protrusions of 0.075mm maximum per side are not included.
 This drawing is subject to change without notice.



12 TAPE AND REEL INFORMATION REEL DIMENSIONS

TAPE DIMENSION



NOTE: The picture is only for reference. Please make the object as the standard.

Package Type	Reel Diameter	Reel Width(mm)	A0 (mm)	B0 (mm)	K0 (mm)	P0 (mm)	P1 (mm)	P2 (mm)	W (mm)	Pin1 Quadrant
SOT23-5	7"	9.5	3.20	3.20	1.40	4.0	4.0	2.0	8.0	Q3
SOP8	13"	12.4	6.40	5.40	2.10	4.0	8.0	2.0	12.0	Q1
MSOP8	13"	12.4	5.20	3.30	1.50	4.0	8.0	2.0	12.0	Q1
TSSOP8	13"	12.4	6.90	3.45	1.65	4.0	8.0	2.0	12.0	Q1
SOP14	13"	16.4	6.60	9.30	2.10	4.0	8.0	2.0	16.0	Q1
TSSOP14	13"	12.4	6.95	5.60	1.20	4.0	8.0	2.0	12.0	Q1
SC70-5	7"	9.5	2.25	2.55	1.20	4.0	4.0	2.0	8.0	Q3
DFN2X2-8	7"	9.5	2.30	2.30	1.10	4.0	4.0	2.0	8.0	Q2

KEY PARAMETER LIST OF TAPE AND REEL

NOTE:

1. All dimensions are nominal.

2. Plastic or metal protrusions of 0.15mm maximum per side are not included.



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