



10MHz, Rail-to-Rail I/O CMOS Operational Amplifier

1 FEATURES

- HIGH GAIN BANDWIDTH:10MHz
- RAIL-TO-RAIL INPUT AND OUTPUT ±0.5mV Typical Vos
- INPUT VOLTAGE RANGE: -0.1V to +5.6V with Vs = 5.5V
- SUPPLY RANGE: +2.5V to +5.5V
- SHUTDOWN: RS721S/RS722S
- SPECIFIED UP TO +125°C
- Micro SIZE PACKAGES: SOT23-5, SOT23-6

2 APPLICATIONS

- SENSORS
- PHOTODIODE AMPLIFICATION
- ACTIVE FILTERS
- TEST EQUIPMENT
- DRIVING A/D CONVERTERS

3 DESCRIPTIONS

The RS72X families of products offer low voltage operation and rail-to-rail input and output, as well as excellent speed/power consumption ratio, providing an excellent bandwidth (10MHz) and slew rate of 7V/us. The op-amps are unity gain stable and feature an ultra-low input bias current.

The devices are ideal for sensor interfaces, active filters and portable applications. The RS721S, RS722S include a shutdown mode. Under logic control, the amplifiers can be switched from normal operation to a standby current that is less than 1uA. The RS72X families of operational amplifiers are specified at the full temperature range of -40°C to +125°C under single or dual power supplies of 2.5V to 5.5V.

Device Information (1)

PART NUMBER	PACKAGE	BODY SIZE(NOM)
	SOT23-5	2.90mm×1.60mm
	SOT23-6	2.90mm×1.60mm
RS721	DFN2X2-6	2.00mm×2.00mm
	SOP8	4.90mm×3.90mm
	MSOP8	3.00mm×3.00mm
	SOP8	4.90mm×3.90mm
	MSOP8	3.00mm×3.00mm
	MSOP10	3.00mm×3.00mm
RS722	DFN2X2-8	2.00mm×2.00mm
	DFN3X3-8 (0303X0.75-0.65)	3.00mm×3.00mm
	DFN3X3-8 (0303X0.75-0.5)	3.00mm×3.00mm
RS724	SOP14	8.65mm×3.90mm
K3/24	TSSOP14	5.00mm×4.40mm

⁽¹⁾ For all available packages, see the orderable addendum at the end of the data sheet.



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4 Revision History

Note: Page numbers for previous revisions may different from page numbers in the current version.

VERSION	Change Date	Change Item
C.1	2021/06/16	Add TDFN3x3-8L(0303x0.75-0.5) package
C.2	2022/06/14	Add TDFN2X2-6L PACKAGE/ORDERING INFORMATION
C.3	2023/09/14	Add MSL on Page 8@RevC.2
C.3.1	2024/03/01	Modify packaging naming



5 PACKAGE/ORDERING INFORMATION (1)

Orderable Device	Package Type	Pin	Channel	Op Temp(°C)	Device Marking ⁽²⁾	MSL ⁽³⁾	Package Qty
RS721XF	SOT23-5	5	1	-40°C ~125°C	721	MSL3	Tape and Reel,3000
RS721BXF	SOT23-5	5	1	-40°C ~125°C	721B	MSL3	Tape and Reel,3000
RS721XK	SOP8	8	1	-40°C ~125°C	RS721	MSL3	Tape and Reel,4000
RS721XM	MSOP8	8	1	-40°C ~125°C	RS721	MSL3	Tape and Reel,4000
RS721XTDE6	DFN2X2-6	6	1	-40°C ~125°C	721	MSL3	Tape and Reel,3000
RS721SXK	SOP8	8	1	-40°C ~125°C	RS721S	MSL3	Tape and Reel,4000
RS721SXH	SOT23-6	6	1	-40°C ~125°C	721S	MSL3	Tape and Reel,3000
RS722XK	SOP8	8	2	-40°C ~125°C	RS722	MSL3	Tape and Reel,4000
RS722XM	MSOP8	8	2	-40°C ~125°C	RS722	MSL3	Tape and Reel,4000
RS722XTDE8	DFN2X2-8	8	2	-40°C ~125°C	722	MSL3	Tape and Reel,3000
RS722XTDC8	DFN3X3-8 (0303X0.75-0.65)	8	2	-40°C ~125°C	RS722	MSL3	Tape and Reel,5000
RS722XTDC8-B	DFN3X3-8 (0303X0.75-0.5)	8	2	-40°C ~125°C	RS722B	MSL3	Tape and Reel,5000
RS722SXN	MSOP10	10	2	-40°C ~125°C	RS722S	MSL3	Tape and Reel,4000
RS724XP	SOP14	14	4	-40°C ~125°C	RS724	MSL3	Tape and Reel,4000
RS724XQ	TSSOP14	14	4	-40°C ~125°C	RS724	MSL3	Tape and Reel,4000

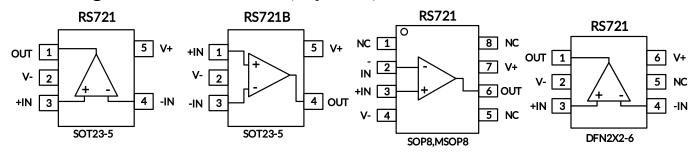
⁽¹⁾ This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the right-hand navigation.

⁽²⁾ There may be additional marking, which relates to the lot trace code information(data code and vendor code), the logo or the environmental category on the device.

⁽³⁾ MSL, The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications.



6 Pin Configuration and Functions (Top View)

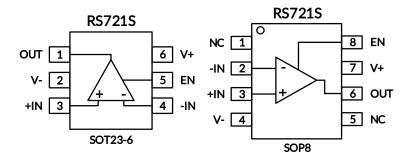


Pin Description

			PIN			
NAME	RS721	RS721B	RS721	RS721	I/O (1)	DESCRIPTION
	SOT23-5	SOT23-5	SOP8/MSOP8	DFN2X2-6		
-IN	4	3	2	4	I	Negative (inverting) input
+IN	3	1	3	3	I	Positive (noninverting) input
NC (2)	-	-	1,5,8	5	-	No internal connection (can be left floating)
OUT	1	4	6	1	0	Output
V-	2	2	4	2	-	Negative (lowest) power supply
V+	5	5	7	6	-	Positive (highest) power supply

⁽¹⁾ I = Input, O = Output.

⁽²⁾ There is no internal connection. Typically, GND is the recommended connection to a heat spreading plane.



Pin Description

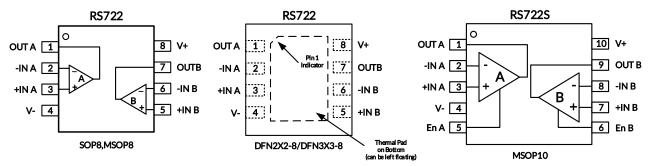
	_	PIN					
NAME	ı	RS721S	I/O (1)	DESCRIPTION			
	SOT23-6 SOP8						
-IN	4	2	I	Inverting input			
+IN	3	3	I	Noninverting input			
OUT	1	6	0	Output			
EN	5	8	I	Enable pin. This pin turns the regulator on or off. Low = disabled, high = normal operation (pin must be driven)			
NC (2)		1,5	-	No internal connection (can be left floating)			
V-	2	2 4 -		Negative (lowest) power supply			
V+	6	7	-	Positive (highest) power supply			

⁽¹⁾ I = Input, O = Output.

⁽²⁾ There is no internal connection. Typically, GND is the recommended connection to a heat spreading plane.



Pin Configuration and Functions (Top View)



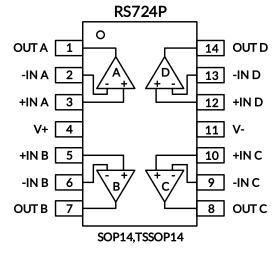
Pin Description

	PIN					
NAME	RS722	RS722S	I/O (1)	DESCRIPTION		
	SOP8/MSOP8/ DFN2X2-8/DFN3X3-8	MSOP10				
-INA	2	2	I	Inverting input, channel A		
+INA	3	3	I	Noninverting input, channel A		
-INB	6	8	I	Inverting input, channel B		
+INB	5	7	I	Noninverting input, channel B		
OUTA	1	1	0	Output, channel A		
OUTB	7	9	0	Output, channel B		
EnA	-	5	ı	Enable pin, channel A. This pin turns the regulator on or off. Low = disabled, high = normal operation (pin must be driven)		
EnB	-	6	I	Enable pin, channel B. This pin turns the regulator on or off. Low = disabled, high = normal operation (pin must be driven)		
V-	4	4	-	Negative (lowest) power supply		
V+	8	10	-	Positive (highest) power supply		
-	Thermal Pad	-	-	Connect thermal pad to V-		

⁽¹⁾ I = Input, O = Output.



Pin Configuration and Functions (Top View)



Pin Description

	PIN	. (-) (1)	
NAME	SOP14/TSSOP14	I/O (1)	DESCRIPTION
-INA	2	I	Inverting input, channel A
+INA	3	1	Noninverting input, channel A
-INB	6	I	Inverting input, channel B
+INB	5	1	Noninverting input, channel B
-INC	9	1	Inverting input, channel C
+INC	10	1	Noninverting input, channel C
-IND	13	1	Inverting input, channel D
+IND	12	I	Noninverting input, channel D
OUTA	1	0	Output, channel A
OUTB	7	0	Output, channel B
OUTC	8	0	Output, channel C
OUTD	14	0	Output, channel D
V-	11	-	Negative (lowest) power supply
V+	4	-	Positive (highest) power supply

⁽¹⁾ I = Input, O = Output.



7 SPECIFICATIONS

7.1 Absolute Maximum Ratings

Over operating free-air temperature range (unless otherwise noted) (1)

			MIN	MAX	UNIT
	Supply, V _S =(V+) - (V-)			7	
Voltage	Signal input pin ⁽²⁾		(V-)-0.5	(V+) +0.5	V
	Signal output pin (3)		(V-)-0.5	(V+) +0.5	
Current	Signal input pin (2)	-10	10	mA	
Current	Signal output pin ⁽³⁾		-100	100	mA
	Output short-circuits (4)		Cont	7 (V+) +0.5 (V+) +0.5 10 100 100 nuous 230 110 170 105 90 230 200 45 80 125 150	
		SOT23-5		230	
		SOP8		110	
		MSOP8		170	
		SOP14		105	
θ_{JA}	Package thermal impedance (5)	TSSOP14		90	°C/W
		SOT23-6		230	
		MSOP10		200	
		DFN3X3-8		45	
		DFN2X2-8		80	
	Operating range, T _A	-40	125		
Temperature	Junction, T _J ⁽⁶⁾		-40	150	°C
	Storage, T _{stg}		-65	7 (V+) +0.5 (V+) +0.5 10 100 100 nuous 230 110 170 105 90 230 200 45 80 125 150	

⁽¹⁾ Stresses above these ratings may cause permanent damage. Exposure to absolute maximum conditions for extended periods may degrade device reliability. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those specified is not implied.

- (4) Short-circuit to ground, one amplifier per package.
- (5) The package thermal impedance is calculated in accordance with JESD-51.
- (6) The maximum power dissipation is a function of $T_{J(MAX)}$, $R_{\theta JA}$, and T_A . The maximum allowable power dissipation at any ambient temperature is $P_D = (T_{J(MAX)} T_A) / R_{\theta JA}$. All numbers apply for packages soldered directly onto a PCB.

7.2 ESD Ratings

The following ESD information is provided for handling of ESD-sensitive devices in an ESD protected area only.

			VALUE	UNIT
V _(ESD)	Electrostatic discharge	Human-body model (HBM)	±5000	V
V (ESD)	Electrostatic discharge	Machine Model (MM)	±400	\ \ \



ESD SENSITIVITY CAUTION

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

⁽²⁾ Input terminals are diode-clamped to the power-supply rails. Input signals that can swing more than 0.5V beyond the supply rails should be current-limited to 10mA or less.

⁽³⁾ Output terminals are diode-clamped to the power-supply rails. Output signals that can swing more than 0.5V beyond the supply rails should be current-limited to ±100mA or less.



7.3 Recommended Operating ConditionsOver operating free-air temperature range (unless otherwise noted)

		MIN	МОМ	MAX	UNIT
Supply voltage Vs= (V+) - (V/)	Single-supply	2.5		5.5	V
Supply voltage, Vs= (V+) - (V-)	Dual-supply	±1.25		±2.75	v



7.4 ELECTRICAL CHARACTERISTICS

(At T_A = +25°C, V_S =5V, R_L = 10k Ω connected to V_S /2, and V_{OUT} = V_S /2, V_{CM} = V_S /2, Full $^{(9)}$ = -40°C to +125°C, unless otherwise noted.) $^{(1)}$

POWER SUPPLY Vs		PARAMETER	CONDITIONS	T,		RS721, F	s, RS722S RS722, RS	
VS Operating Voltage Range 25°C 2.5 5.5 IQ Quiescent Current/Amplifier 25°C 1.15 1.4 n PSRR Ratio Power-Supply Rejection Vs=2.5V to 5.5V,					MIN ⁽²⁾	TYP ⁽³⁾	5.5 1.4 2.5 ±10 ±10 5.6	UNIT
IQ	POWER S	UPPLY	T					
PSRR Power-Supply Rejection Nose N	Vs	Operating Voltage Range		25°C	2.5		5.5	V
NPUT	IQ (Quiescent Current/Amplifier		25°C		1.15	1.4	mA
Natio Vex=(V-)+0.5V Full 65 65 65	DSDD	Power-Supply Rejection		25°C	75	85		dB
Vos Input Offset Voltage Vcm=Vs/2 25°C -2.5		Ratio	V _{CM} =(V-)+0.5V	Full	65			ав
Vos TC			T	T		1	1	
Vos C Drift			V _{CM} =V _S /2	25°C	-2.5	±0.5	2.5	mV
Input Offset Current (5)			V _{CM} =V _S /2	Full		±2.6		μV/°C
Vcm Common-Mode Voltage Range Vs= 5.5V 25°C -0.1 5.6 CMRR Range Vs= 5.5V, Vs= 5.5V, VcM=-0.1V to 4V Full 67 67 Common-Mode Rejection Ratio Rest StV, VcM=-0.1V to 4V Full 67 60 OUTPUT AOL Open-Loop Voltage Gain Rest XΩ, VcM=-0.1V to 4.85V Full 57 7 60 Quently Swing from Rail Rest XΩ, Vc=0.15V to 4.85V Full 25°C 86 95 90 96 96 96 96 90 96	IB I	Input Bias Current (4) (5)		25°C		±1	±10	pА
Common-Mode Rejection Ratio Vs= 5.5V, VcM=-0.1V to 4V Full 67 67 67 67 67 67 67	los	Input Offset Current (5)		25°C		±1	±10	pА
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	vcm i	_	Vs= 5.5V	25°C	-0.1		5.6	V
CMRR Ratio Common-Mode Rejection Ratio V _{CM} =-0.1V to 4V Full 57 67 Common-Mode Rejection Ratio V _{CM} =-0.1V to 5.6V Full 57 Common-Mode Rejection Ratio Common-Mode Rejection Ration			Vs= 5.5V.	25°C	73	85		
$ \begin{array}{ c c c c c c } \hline \text{CMRR} & \text{Ratio} & V_{S=5.5V, V_{CM}=-0.1V to 5.6V} & 25^{\circ}\text{C} & 60 & 75 & & & & & & & & & & & & & & & & & $		Common-Mode Rejection	*	Full	67			
Note	(IVIRR I	-	Vs= 5 5V	25°C	60	75		dB
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$					57			
$AOL \ \ \begin{array}{ c c c c c c }\hline \ \ & AOL \ \ \\ \ \\ \ \ \\ \ \ \\ \ \\ \ \\ \ \\ \ \\ \ \\ \ $	OUTPUT				1 0,			
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$			R _I =2KO	25°C	86	95		
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$		0 1 1/4 6:				1,3		ID
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	AOL	Open-Loop Voltage Gain				96		dB
Output Swing from Rail R _L =10KΩ 25°C 7 n n				Full	66			
Lout Output Short-Circuit Current 25°C		Output Swing from Rail		25°C				mV
FREQUENCY RESPONSE SR Slew Rate (8) 25°C 7 V V V V V V V V V		Output Short-Circuit Current	KF=10K77			/		
SR Slew Rate (8) 25°C 7 V GBP Gain-Bandwidth Product 25°C 10 M PM Phase Margin (5) 25°C 62 — ts Setting Time,0.1% 0.2 — — Overload Recovery Time V _{IN} ·Gain≥Vs 0.35 — NOISE en Input Voltage Noise Density f = 1KHz 25°C 9.5 nV/ ENABLE/SHUTDOWN (RS721S, RS722S) F 10KHz 25°C <1				25°C		±70		mA
GBP Gain-Bandwidth Product 25°C 10 M PM Phase Margin (5) 25°C 62	FREQUEN	ICY RESPONSE						
PM Phase Margin (5) 25°C 62 ts Setting Time, 0.1% 0.2 1 Overload Recovery Time V _{IN} -Gain≥Vs 0.35 1 NOISE en Input Voltage Noise Density f = 1KHz 25°C 9.5 nV/obstance ENABLE/SHUTDOWN (RS721S, RS722S) f = 10KHz 25°C 6.5 nV/obstance LQ(OFF) Supply Current in Shutdown 25°C 3 1 toff 25°C 3 1 ton 25°C 20 1 VL Shut Down 25°C V- (V-)	SR :	Slew Rate ⁽⁸⁾		25°C		7		V/µs
ts Setting Time,0.1% 0.2 I Overload Recovery Time V _{IN} ·Gain≥Vs 0.35 I NOISE en Input Voltage Noise Density f = 1KHz 25°C 9.5 nV/ f = 10KHz 25°C 6.5 nV/ ENABLE/SHUTDOWN (RS721S, RS722S) IQ(OFF) Supply Current in Shutdown 25°C <1	GBP	Gain-Bandwidth Product		25°C		10		MHz
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	PM I	Phase Margin (5)		25°C		62		0
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	ts :	Setting Time,0.1%				0.2		μs
NOISE en Input Voltage Noise Density f = 1KHz 25°C 9.5 nV/y f = 10KHz 25°C 6.5 nV/y ENABLE/SHUTDOWN (RS721S, RS722S) IQ(OFF) Supply Current in Shutdown 25°C 41 14 toff 25°C 3 14 toN 25°C 20 14 VL Shut Down 25°C V- 40.8		Overload Recovery Time	V _{IN} •Gain≥V _S			0.35		μs
ENABLE/SHUTDOWN (RS721S, RS722S) f = 10KHz 25°C 6.5 nV/L Iq(off) Supply Current in Shutdown 25°C <1		,			1	1		
ENABLE/SHUTDOWN (RS721S, RS722S) f = 10KHz 25°C 6.5 nV/L Iq(off) Supply Current in Shutdown 25°C <1			f = 1KHz	25°C		9.5		nV/√Hz
ENABLE/SHUTDOWN (RS721S, RS722S) IQ(OFF) Supply Current in Shutdown 25°C <1	en l	Input Voltage Noise Density		-				nV/√Hz
IQ(OFF) Supply Current in Shutdown 25°C <1 I toff 25°C 3 I toN 25°C 20 I VL Shut Down 25°C V- (V-) (V-) +0.8	ENABLE/S	SHUTDOWN (RS721S, RS722		25 C		0.5		114/ (112
toff 25°C 3 1 toN 25°C 20 1 VL Shut Down 25°C V- (V-) +0.8		•		25°C		<1		μΑ
ton 25°C 20 I V _L Shut Down 25°C V- (V-) +0.8		11 /				-		μs
VL Shut Down 25°C V- (V-) +0.8				_		-		μs
+0.8		Shut Down			V-	20		μs V
V_H Amplifier Is Active 25°C (V-) +2 V+		Amplifier Is Active		25°C	(V-) +2	1		V



- (1) Electrical table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device.
- (2) Limits are 100% production tested at 25°C. Limits over the operating temperature range are ensured through correlations using statistical quality control (SQC) method.
- (3) Typical values represent the most likely parametric norm as determined at the time of characterization. Actual typical values may vary over time and will also depend on the application and configuration.
- (4) Positive current corresponds to current flowing into the device.
- (5) This parameter is ensured by design and/or characterization and is not tested in production.
- (6) The maximum power dissipation is a function of $T_{J(MAX)}$, $R_{\theta JA}$, and T_A . The maximum allowable power dissipation at any ambient temperature is PD = $(T_{J(MAX)} T_A) / R_{\theta JA}$. All numbers apply for packages soldered directly onto a PCB.
- (7) Short circuit test is a momentary test.
- (8) Number specified is the slower of positive and negative slew rates.
- (9) Specified by characterization only.



7.5 TYPICAL CHARACTERISTICS

NOTE: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only.

At T_A = +25°C, V_S =5V, R_L = 10k Ω connected to V_S /2, V_{OUT} = V_S /2, unless otherwise noted.

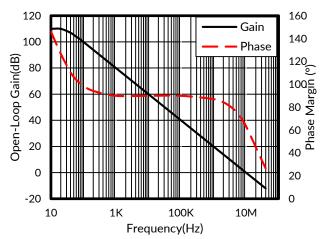


Figure 1. OPEN-LOOP GAIN AND PHASE vs FREQUENCY

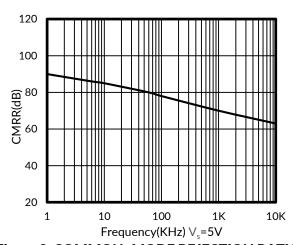


Figure 2. COMMON-MODE REJECTION RATIO vs FREQUENCY

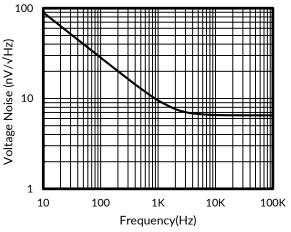


Figure 3. INPUT VOLTAGE NOISE SPECTRAL DENSITY vs FREQUENCY

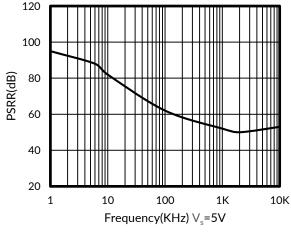


Figure 4. POWER-SUPPLY REJECTION RATIO vs FREQUENCY

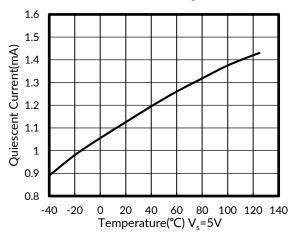


Figure 5. QUIESCENT CURRENT vs TEMPERATURE

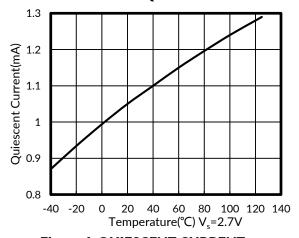


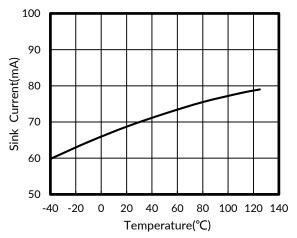
Figure 6. QUIESCENT CURRENT vs TEMPERATURE



TYPICAL CHARACTERISTICS

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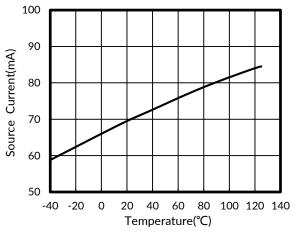
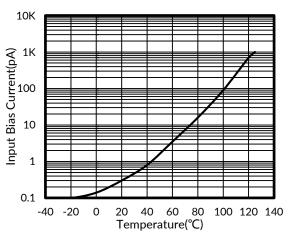


Figure 7. SINK CURRENT vs TEMPERATURE

Figure 8. SOURCE CURRENT vs TEMPERATURE



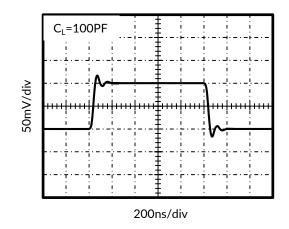
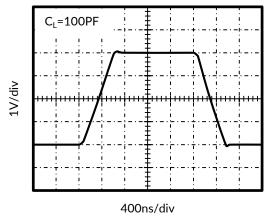


Figure 9. INPUT BIAS CURRENT vs TEMPERATURE

Figure 10. SMALL-SIGNAL STEP RESPONSE



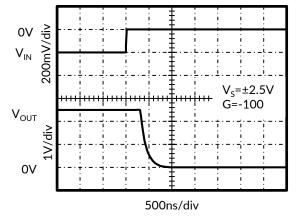


Figure 11. LARGE-SIGNAL STEP RESPONSE

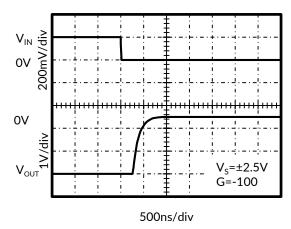
Figure 12. POSITIVE OVERLOAD RECOVERY



TYPICAL CHARACTERISTICS

NOTE: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only.

At T_A = +25°C, V_S =5V, R_L = 10k Ω connected to V_S /2, V_{OUT} = V_S /2, unless otherwise noted.



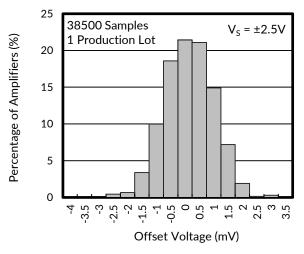


Figure 13. NEGATIVE OVERLOAD RECOVERY

Figure 14. OFFSET VOLTAGE PRODUCTION DISTRIBUTION



8 Detailed Description

8.1 Overview

The RS721, RS722, RS724, RS721S, RS722S are high precision, rail-to-rail operational amplifiers that can be run from a single-supply voltage 2.5V to $5.5V(\pm 1.25V$ to $\pm 2.75V$). Supply voltages higher than 7V(absolute maximum) can permanently damage the amplifier. Rail-to-rail input and output swing significantly increases dynamic range, especially in low-supply applications. Good layout practice mandates use of a 0.1uF capacitor place closely across the supply pins.

8.2 RS721S/RS722S ENABLE FUNCTION

The RS721S/RS722S includes a shutdown mode. Under logic control, the amplifiers can be switched from normal mode to a standby current of 1uA. When the Enable pin is connected to high, the amplifier is active. Connecting Enable low disables the amplifier, and places the amplifier, and places the output in a high-impedance state.

8.3 Phase Reversal Protection

The RS72X family has internal phase-reversal protection. Many op amps exhibit phase reversal when the input is driven beyond the linear common-mode range. This condition is most often encountered in noninverting circuits when the input is driven beyond the specified common-mode voltage range, causing the output to reverse into the opposite rail. The input of the RS72X prevents phase reversal with excessive common-mode voltage. Instead, the appropriate rail limits the output voltage. This performance is shown in figure 15.

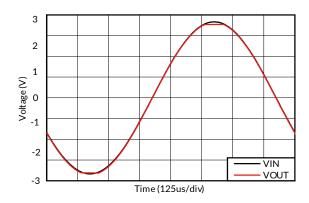


Figure 15. Output Waveform Devoid of Phase Reversal During an Input Overdrive Condition

8.4 EMI Rejection Ratio (EMIRR)

The electromagnetic interference (EMI) rejection ratio, or EMIRR, describes the EMI immunity of operational amplifiers. An adverse effect that is common to many operational amplifiers is a change in the offset voltage as a result of RF signal rectification. An operational amplifier that is more efficient at rejecting this change in offset as a result of EMI has a higher EMIRR and is quantified by a decibel value. Measuring EMIRR can be performed in many ways, but this document provides the EMIRR IN+, which specifically describes the EMIRR performance when the RF signal is applied to the noninverting input pin of the operational amplifier. In general, only the noninverting input is tested for EMIRR for the following three reasons:

- Operational amplifier input pins are known to be the most sensitive to EMI, and typically rectify RF signals better than the supply or output pins.
- The noninverting and inverting operational amplifier inputs have symmetrical physical layouts and exhibit nearly matching EMIRR performance.
- EMIRR is easier to measure on noninverting pins than on other pins because the noninverting input pin can be isolated on a printed-circuit-board (PCB). This isolation allows the RF signal to be applied directly to the noninverting input pin with no complex interactions from other components or connecting PCB traces.



Detailed Description(continued)

The EMIRR IN+ of the RS72X is plotted versus frequency in Figure 16. If available, any dual and quad operational amplifier device versions have approximately identical EMIRR IN+ performance. The RS72X unity-gain bandwidth is 10MHz. EMIRR performance below this frequency denotes interfering signals that fall within the operational amplifier bandwidth.

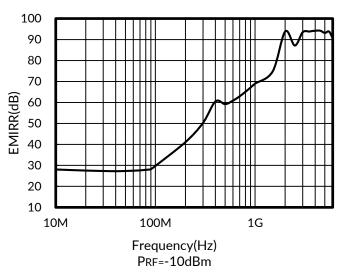


Figure 16. RS72X EMIRR vs Frequency

8.5 EMIRR IN+ Test Configuration

Figure 17 shows the circuit configuration for testing the EMIRR IN+. An RF source is connected to the operational amplifier noninverting input pin using a transmission line. The operational amplifier is configured in a unity-gain buffer topology with the output connected to a low-pass filter (LPF) and a digital multimeter (DMM). A large impedance mismatch at the operational amplifier input causes a voltage reflection; however, this effect is characterized and accounted for when determining the EMIRR IN+. The resulting dc offset voltage is sampled and measured by the multimeter. The LPF isolates the multimeter from residual RF signals that can interfere with multimeter accuracy.

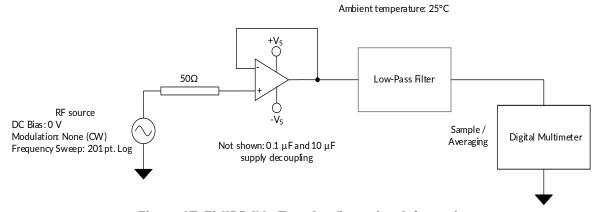


Figure 17. EMIRR IN+ Test Configuration Schematic

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9 Application and Implementation

Information in the following applications sections is not part of the RUNIC component specification, and RUNIC does not warrant its accuracy or completeness. RUNIC's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

9.1 APPLICATION NOTE

The RS72X are high precision, rail-to-rail operational amplifiers that can be run from a single-supply voltage 2.5V to 5.5V ($\pm 1.25V$ to $\pm 2.75V$). Supply voltages higher than 7V (absolute maximum) can permanently damage the amplifier. Rail-to-rail input and output swing significantly increases dynamic range, especially in low-supply applications. Good layout practice mandates use of a 0.1uF capacitor place closely across the supply pins.

Typical Applications 9.2 25-kHz Low-pass Filter

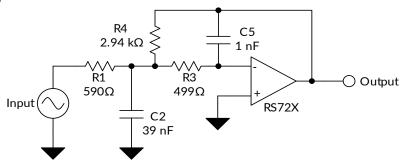


Figure 18. 25-kHz Low-Pass Filter

9.3 Design Requirements

Low-pass filters are commonly employed in signal processing applications to reduce noise and prevent aliasing. The RS72X devices are ideally suited to construct high-speed, high-precision active filters. Figure 18 shows a second-order, low-pass filter commonly encountered in signal processing applications.

Use the following parameters for this design example:

- Gain = 5 V/V (inverting gain)
- Low-pass cutoff frequency = 25 kHz
- Second-order Chebyshev filter response with 3-dB gain peaking in the passband

9.4 Detailed Design Procedure

The infinite-gain multiple-feedback circuit for a low-pass network function is shown in Figure 18. Use Equation 1 to calculate the voltage transfer function.

$$\frac{\text{Output}}{\text{Input}}(s) = \frac{-1/R_1R_3C_2C_5}{s^2 + (s/C_2)(1/R_1 + 1/R_3 + 1/R_4) + 1/R_3R_4C_2C_5}$$
(1)

This circuit produces a signal inversion. For this circuit, the gain at dc and the low-pass cutoff frequency are calculated by Equation 2:

Gain =
$$\frac{R_4}{R_1}$$

 $f_C = \frac{1}{2\pi} \sqrt{(1/R_3 R_4 C_2 C_5)}$

(2)



9.5 Application Curve

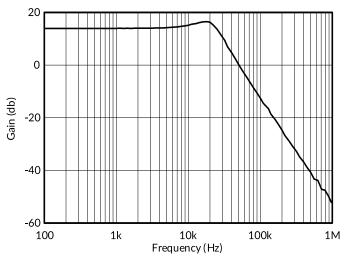


Figure 19. Low-pass filter transfer function



10 Layout

10.1 Layout Guidelines

Attention to good layout practices is always recommended. Keep traces short. When possible, use a PCB ground plane with surface-mount components placed as close to the device pins as possible. Place a 0.1uF capacitor closely across the supply pins.

These guidelines should be applied throughout the analog circuit to improve performance and provide benefits such as reducing the EMI susceptibility.

10.2 Layout Example

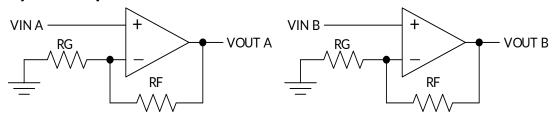


Figure 20. Schematic Representation

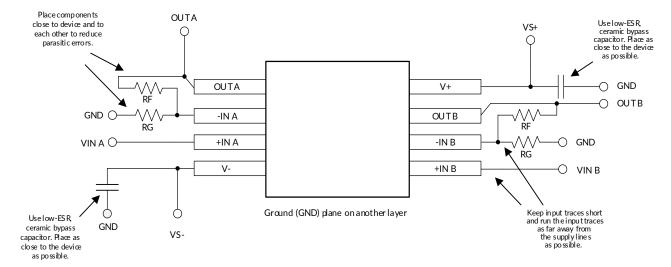
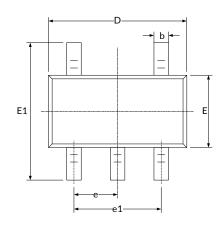
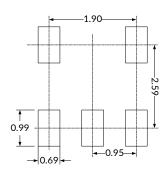


Figure 21. Layout Example

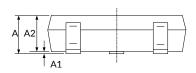


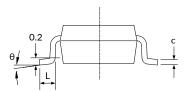
11 PACKAGE OUTLINE DIMENSIONS SOT23-5 (3)





RECOMMENDED LAND PATTERN (Unit: mm)



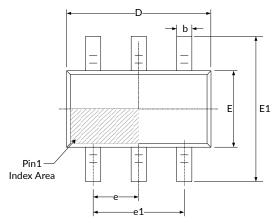


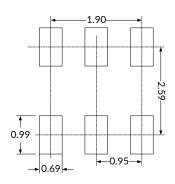
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Symbol	Min	Min Max		Max	
A (1)	1.050	1.250	0.041	0.049	
A1	0.000	0.100	0.000	0.004	
A2	1.050	1.150	0.041	0.045	
b	0.300	0.500	0.012	0.020	
С	0.100	0.200	0.004	0.008	
D (1)	2.820	3.020	0.111	0.119	
E (1)	1.500	1.700 0.059		0.067	
E1	2.650	2.950	0.104	0.116	
е	0.950(BSC) ⁽²⁾	0.037(BSC) (2)		
e1	1.800	2.000	0.071	0.079	
L	0.300	0.600	0.012	0.024	
θ	0°	8°	0°	8°	

- 1. Plastic or metal protrusions of 0.15mm maximum per side are not included.
- BSC (Basic Spacing between Centers), "Basic" spacing is nominal.
 This drawing is subject to change without notice.

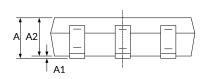


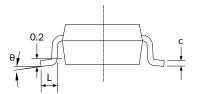
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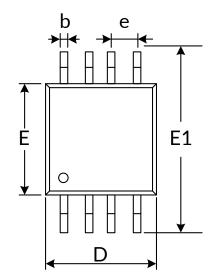


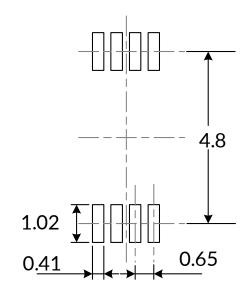
Complete	Dimensions I	n Millimeters	Dimensions In Inches		
Symbol	Min	Min Max		Max	
A (1)	1.050	1.250	0.041	0.049	
A1	0.000	0.100	0.000	0.004	
A2	1.050	1.150	0.041	0.045	
b	0.300	0.500	0.012	0.020	
С	0.100	0.200	0.004	0.008	
D (1)	2.820	3.020	0.111	0.119	
E (1)	1.500	1.700	0.059	0.067	
E1	2.650	2.950	0.104	0.116	
е	0.950(BSC) (2)	0.037(BSC) ⁽²⁾	
e1	1.800	2.000	0.071	0.079	
L	0.300	0.600	0.012	0.024	
θ	0°	8°	0° 8°		

- 1. Plastic or metal protrusions of 0.15mm maximum per side are not included.
- BSC (Basic Spacing between Centers), "Basic" spacing is nominal.
 This drawing is subject to change without notice.

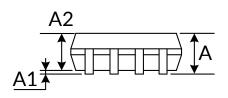


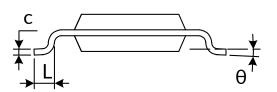
MSOP8 (3)





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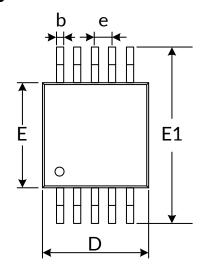


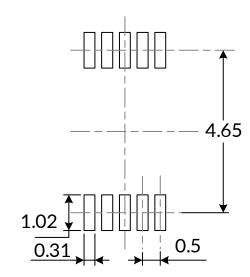
Complete I	Dimensions I	n Millimeters	Dimensions In Inches		
Symbol	Min	Max	Min	Max	
A ⁽¹⁾	0.820	1.100	0.032	0.043	
A1	0.020	0.150	0.001	0.006	
A2	0.750	0.950	0.030	0.037	
b	0.250	0.380	0.010	0.015	
С	0.090	0.230	0.004	0.009	
D (1)	2.900	3.100	0.114	0.122	
е	0.650(BSC) (2)	0.026(BSC) (2)		
E (1)	2.900	3.100	0.114	0.122	
E1	4.750	5.050	0.187	0.199	
L	0.400	0.800	0.016	0.031	
θ	0°	6°	0°	6°	

- 1. Plastic or metal protrusions of 0.15mm maximum per side are not included.
- 2. BSC (Basic Spacing between Centers), "Basic" spacing is nominal.
- 3. This drawing is subject to change without notice.

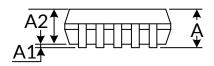


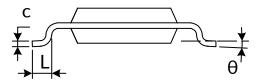
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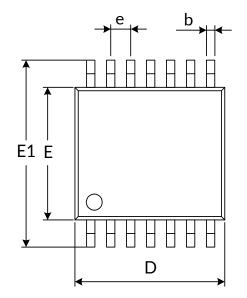


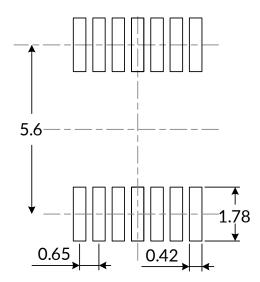
Complete	Dimensions I	n Millimeters	Dimension	s In Inches	
Symbol	Min	Max	Min	Max	
A (1)	0.820	1.100	0.032	0.043	
A1	0.020	0.150	0.001	0.006	
A2	0.750	0.950	0.030	0.037	
b	0.180	0.280	0.007	0.011	
С	0.090	0.230	0.004	0.009	
D ⁽¹⁾	2.900	3.100	0.114	0.122	
е	0.50(E	3SC) ⁽²⁾	0.020(BSC) ⁽²⁾		
E ⁽¹⁾	2.900	3.100	0.114	0.122	
E1	4.750	5.050	0.187	0.199	
L	0.400	0.800	0.016	0.031	
θ	θ 0° 6°		0°	6°	

- Plastic or metal protrusions of 0.15mm maximum per side are not included.
 BSC (Basic Spacing between Centers), "Basic" spacing is nominal.
- 3. This drawing is subject to change without notice.

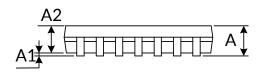


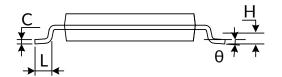
TSSOP14 (3)





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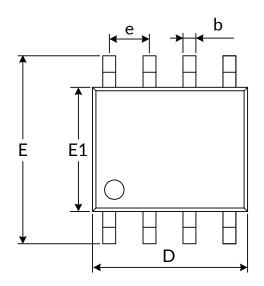


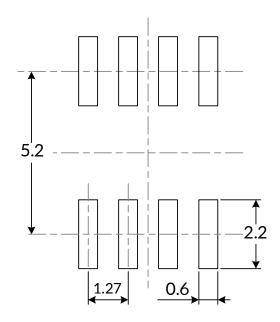
Complete	Dimensions I	n Millimeters	Dimensions In Inches		
Symbol	Min	Max	Min	Max	
A (1)		1.200		0.047	
A1	0.050	0.150	0.002	0.006	
A2	0.800	1.050	0.031	0.041	
b	0.190	0.300	0.007	0.012	
С	0.090	0.200	0.004	0.008	
D ⁽¹⁾	4.860	5.100	0.191	0.201	
E ⁽¹⁾	4.300	4.500	0.169	0.177	
E1	6.250	6.550	0.246	0.258	
е	0.650(BSC) (2)	0.026(BSC) (2)		
L	0.500	0.700	0.020	0.028	
Н	0.25(TYP)		0.01(TYP)		
θ	1°	7°	1° 7°		

- Plastic or metal protrusions of 0.15mm maximum per side are not included.
 BSC (Basic Spacing between Centers), "Basic" spacing is nominal.
- 3. This drawing is subject to change without notice.

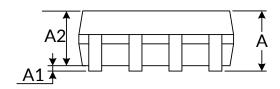


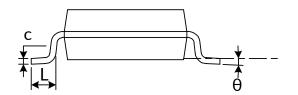
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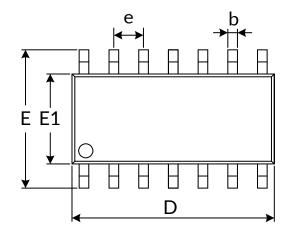


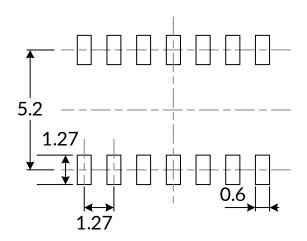
S. mahal	Dimensions I	n Millimeters	Dimension	s In Inches
Symbol	Min	Min Max		Max
A (1)	1.350	1.750	0.053	0.069
A1	0.100	0.250	0.004	0.010
A2	1.350 1.550 0.053		0.053	0.061
b	0.330	0.510	0.013	0.020
С	0.170	0.250	0.007	0.010
D ⁽¹⁾	4.800	5.000	0.189	0.197
е	1.270(BSC) (2)	0.050(BSC) (2)	
Е	5.800	6.200	0.228	0.244
E1 ⁽¹⁾	3.800	4.000	0.150	0.157
L	0.400	1.270	0.016	0.050
θ	θ 0° 8°		0°	8°

- Plastic or metal protrusions of 0.15mm maximum per side are not included.
 BSC (Basic Spacing between Centers), "Basic" spacing is nominal.
 This drawing is subject to change without notice.

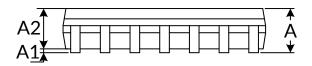


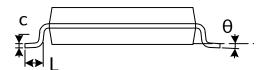
SOP14 (3)





RECOMMENDED LAND PATTERN (Unit: mm)



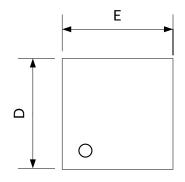


Symphol	Dimensions I	n Millimeters	Dimension	s In Inches	
Symbol	Min	Min Max		Max	
A (1)	1.350	1.750	0.053	0.069	
A1	0.100	0.250	0.004	0.010	
A2	1.350 1.550		0.053	0.061	
b	0.310	0.510	0.012	0.020	
С	0.100	0.250	0.004	0.010	
D ⁽¹⁾	8.450	8.850	0.333	0.348	
e	1.270(BSC) (2)	0.050(BSC) (2)		
Е	5.800	6.200	0.228	0.244	
E1 ⁽¹⁾	3.800	4.000	0.150	0.157	
L	0.400	1.270	0.016	0.050	
θ	θ 0° 8°		0°	8°	

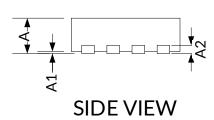
- 1. Plastic or metal protrusions of 0.15mm maximum per side are not included.
- 2. BSC (Basic Spacing between Centers), "Basic" spacing is nominal.
 3. This drawing is subject to change without notice.

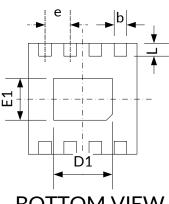


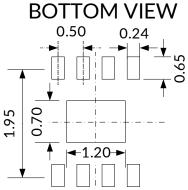
DFN2X2-8 (2)



TOP VIEW







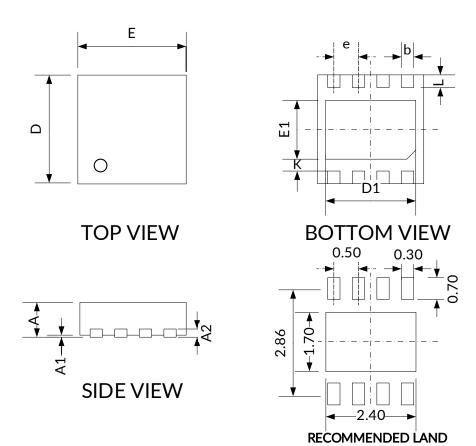
RECOMMENDED LAND PATTERN (Unit: mm)

Symbol	Dimensions I	n Millimeters	Dimension	s In Inches
Symbol	Min	Max	Min	Max
A ⁽¹⁾	0.700	0.800	0.028	0.031
A1	0.000	0.050	0.000	0.002
A2	0.203(TYP)		0.008(TYP)	
b	0.180	0.300	0.007	0.012
D ⁽¹⁾	1.900	2.100	0.075	0.083
D1	1.100	1.300	0.043	0.051
E (1)	1.900	2.100	0.075	0.083
E1	0.600	0.800	0.024 0.031	
е	0.500(TYP)		0.020(TYP)	
L	0.250	0.450	450 0.010	

- Plastic or metal protrusions of 0.075mm maximum per side are not included.
 This drawing is subject to change without notice.



DFN3X3-8 (0303X0.75-0.5) (3)



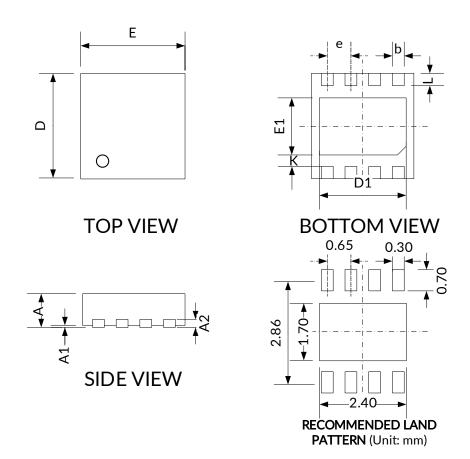
Sand hal	Dimensions I	n Millimeters	Dimension	s In Inches	
Symbol	Min	Max	Min	Max	
A ⁽¹⁾	0.700	0.800	0.028	0.031	
A1	0.000	0.050	0.000	0.002	
A2	0.203	REF (2)	0.008	REF (2)	
b	0.200	0.300	0.008	0.012	
D ⁽¹⁾	2.900	3.100	0.114	0.122	
D1	2.350	2.550	0.093	0.100	
E ⁽¹⁾	2.900	3.100	0.114	0.122	
E1	1.500	1.700	0.059	0.067	
е	0.500	0.500 TYP) TYP	
L	0.350	0.450	0.014	0.018	
К	0.300 REF ⁽²⁾		0.012 REF ⁽²⁾		

PATTERN (Unit: mm)

- 1. Plastic or metal protrusions of 0.075mm maximum per side are not included.
- 2. REF is the abbreviation for Reference.3. This drawing is subject to change without notice.



DFN3X3-8 (0303X0.75-0.65) (3)

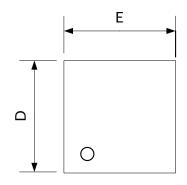


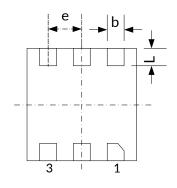
Complete	Dimensions I	n Millimeters	Dimensions In Inches		
Symbol	Min	Max	Min	Max	
A ⁽¹⁾	0.700	0.800	0.028	0.031	
A1	0.000	0.050	0.000	0.002	
A2	0.203	REF (2)	0.008 REF ⁽²⁾		
b	0.200	0.300	0.008	0.012	
D (1)	2.900	3.100	0.114	0.122	
D1	2.250	2.350	0.089	0.093	
E (1)	2.900	3.100	0.114	0.122	
E1	1.450	1.550	0.057	0.061	
е	0.650 TYP		0.026 TYP		
L	0.425	0.525	0.017	0.021	
К	0.200 REF ⁽²⁾		0.008 REF ⁽²⁾		

- Plastic or metal protrusions of 0.075mm maximum per side are not included.
 REF is the abbreviation for Reference.
- 3. This drawing is subject to change without notice.

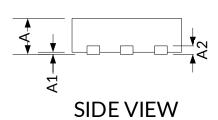


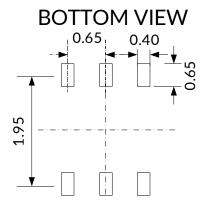
DFN2X2-6 (2)





TOP VIEW





RECOMMENDED LAND PATTERN (Unit: mm)

S.m.h.al	Dimensions I	n Millimeters	Dimensions In Inches		
Symbol	Min	Max	Min	Max	
A ⁽¹⁾	0.700	0.800	0.028	0.031	
A1	0.000 0.050		0.000	0.002	
A2	0.203	B(TYP)	0.008(TYP)		
b	0.250	0.350	0.010	0.012	
D (1)	1.900	2.100	0.075	0.083	
E ⁽¹⁾	1.900	2.100	0.075	0.083	
е	0.650(TYP)		0.026(TYP)		
L	0.250	0.400	0.010	0.018	

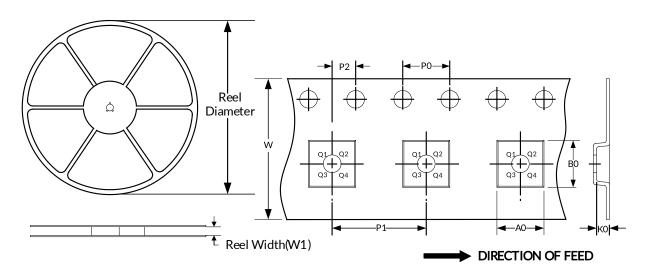
- 1. Plastic or metal protrusions of 0.075mm maximum per side are not included.
- 2. This drawing is subject to change without notice.



12 TAPE AND REEL INFORMATION

REEL DIMENSIONS

TAPE DIMENSION



NOTE: The picture is only for reference. Please make the object as the standard.

KEY PARAMETER LIST OF TAPE AND REEL

	11740 UNE LENGTOT THE ENGLISHED									
Package Type	Reel Diameter	Reel Width (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P0 (mm)	P1 (mm)	P2 (mm)	W (mm)	Pin1 Quadrant
SOT23-5	7"	9.5	3.20	3.20	1.40	4.0	4.0	2.0	8.0	Q3
SOT23-6	7"	9.5	3.17	3.23	1.37	4.0	4.0	2.0	8.0	Q3
SOP8	13"	12.4	6.40	5.40	2.10	4.0	8.0	2.0	12.0	Q1
MSOP8	13"	12.4	5.20	3.30	1.50	4.0	8.0	2.0	12.0	Q1
MSOP10	13"	12.4	5.20	3.30	1.20	4.0	8.0	2.0	12.0	Q1
SOP14	13"	16.4	6.60	9.30	2.10	4.0	8.0	2.0	16.0	Q1
TSSOP14	13"	12.4	6.95	5.60	1.20	4.0	8.0	2.0	12.0	Q1
DFN2X2-6	7"	9.5	2.30	2.30	1.10	4.0	4.0	2.0	8.0	Q1
DFN2x2-8	7"	9.5	2.30	2.30	1.10	4.0	4.0	2.0	8.0	Q2
DFN3X3-8 (0303X0.75-0.5)	13"	12.4	3.35	3.35	1.13	4.0	8.0	2.0	12.0	Q1
DFN3X3-8 (0303X0.75-0.65)	13"	12.4	3.35	3.35	1.13	4.0	8.0	2.0	12.0	Q1

- 1. All dimensions are nominal.
- 2. Plastic or metal protrusions of 0.15mm maximum per side are not included.



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