

Nano-Power, CMOS Input, RRIO, Push-Pull Output Comparator

1 FEATURES

- **Low supply current**
400nA (TYP) at $V_S = 1.4V$
- **Low input offset voltage: $V_{os(max)} = \pm 3mV$**
- **Rail-to-Rail Input and output**
- **SUPPLY RANGE: +1.4V to +5.5V**
- **SPECIFIED UP TO +125°C**
- **Micro SIZE PACKAGES: SOP8, MSOP8**

2 APPLICATIONS

- **OVERVOLTAGE AND UNDERVOLTAGE DETECTION**
- **MULTIVIBRATORS**
- **OVERCURRENT DETECTION**
- **SYSTEM MONITORING**
- **BATTERY POWERED SYSTEM**

3 DESCRIPTIONS

The RS8905 offers a wide supply range, It is a dual low power comparator with a typical power supply current of 400nA (per channel), and rail-to-rail inputs. All of these features come in industry-standard and extremely small packages, making this device an excellent choice for low-voltage and low-power applications for portable electronics and industrial systems.

Featuring a push-pull output stage, the RS8905 allows for operation with absolute minimum power consumption when driving any capacitive or resistive load.

The devices are ideal for system monitoring, include tablets, portable medical, smart phones. The RS8905 is specified at the full temperature range of $-40^{\circ}C$ to $+125^{\circ}C$ under single power supplies of 1.4V to 5.5V.

Device Information ⁽¹⁾

PART NUMBER	PACKAGE	BODY SIZE (NOM)
RS8905	SOP8	4.90mm×3.90mm
	MSOP8	3.00mm×3.00mm

(1) For all available packages, see the orderable addendum at the end of the data sheet.

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4 REVISION HISTORY

Note: Page numbers for previous revisions may differ from page numbers in the current version.

VERSION	Change Date	Change Item
A.6	2022/4/15	<ol style="list-style-type: none"> 1. Delete the second remark in the Absolute Maximum Ratings in Page 3@A.5 Version. 2. Change Functional Block Diagram 3. Added TAPE AND REEL INFORMATION 4. Change ELECTRICAL CHARACTERISTICS in Page 5@A.5 Version. 5. Change TYPICAL CHARACTERISTICS in Page 6@A.5 Version.
A.6.1	2024/02/23	Modify packaging naming
A.7	2024/03/27	<ol style="list-style-type: none"> 1. Added MSL on Page 5@RevA.6.1 2. Update Package thermal impedance 3. Update PACKAGE note
A.8	2025/03/21	Add Charged-device model (CDM)

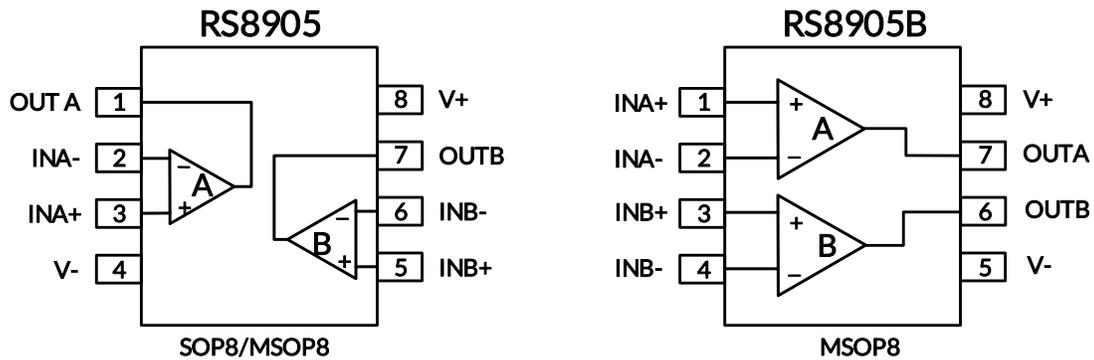
5 PACKAGE/ORDERING INFORMATION ⁽¹⁾

Orderable Device	Package Type	Pin	Channel	Op Temp(°C)	Device Marking ⁽²⁾	MSL ⁽³⁾	Package Qty
RS8905XK	SOP8	8	2	-40°C ~125°C	RS8905	MSL3	Tape and Reel, 4000
RS8905XM	MSOP8	8	2	-40°C ~125°C	RS8905	MSL3	Tape and Reel, 4000
RS8905BXM	MSOP8	8	2	-40°C ~125°C	RS8905B	MSL3	Tape and Reel, 4000

NOTE:

- (1) This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the right-hand navigation.
- (2) There may be additional marking, which relates to the lot trace code information (data code and vendor code), the logo or the environmental category on the device.
- (3) Runic classify the MSL level with using the common preconditioning setting in our assembly factory conforming to the JEDEC industrial standard J-STD-20F, Please align with Runic if your end application is quite critical to the preconditioning setting or if you have special requirement.

6 PIN CONFIGURATION AND FUNCTIONS (TOP VIEW)



Pin Description

NAME	PIN		I/O ⁽¹⁾	DESCRIPTION
	RS8905B	RS8905		
	MSOP8	SOP8/MSOP8		
INA+	1	3	I	Noninverting input A
INA-	2	2	I	Inverting input A
INB+	3	5	I	Noninverting input B
INB-	4	6	I	Inverting input B
V-	5	4	P	Negative (lowest) power supply
OUTB	6	7	O	Output B
OUTA	7	1	O	Output A
V+	8	8	P	Positive (highest) power supply

(1) I=Input, O=Output, P=Power.

7 SPECIFICATIONS

7.1 Absolute Maximum Ratings

Over operating free-air temperature range (unless otherwise noted) ⁽¹⁾

		MIN	MAX	UNIT
Voltage	Supply, $V_S=(V+) - (V-)$		7	V
	Input pin (IN+, IN-)	(V-)-0.5	(V+) +0.5	
	Signal output pin ⁽²⁾	(V-)-0.5	(V+) +0.5	
Current	Signal input pin (IN+, IN-)	-10	10	mA
	Signal output pin ⁽²⁾	-55	55	mA
	Output short-circuit ⁽³⁾	Continuous		
θ_{JA}	Package thermal impedance ⁽⁴⁾	SOP8	110	°C/W
		MSOP8	170	
Temperature	Operating range, T_A	-40	125	°C
	Junction, T_J ⁽⁵⁾	-40	150	
	Storage, T_{stg}	-65	150	

(1) Stresses above these ratings may cause permanent damage. Exposure to absolute maximum conditions for extended periods may degrade device reliability. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those specified is not implied.

(2) Output terminals are diode-clamped to the power-supply rails. Output signals that can swing more than 0.5V beyond the supply rails should be current-limited to ± 55 mA or less.

(3) Short-circuit to ground, one amplifier per package.

(4) The package thermal impedance is calculated in accordance with JESD-51.

(5) The maximum power dissipation is a function of $T_{J(MAX)}$, $R_{\theta JA}$, and T_A . The maximum allowable power dissipation at any ambient temperature is $P_D = (T_{J(MAX)} - T_A) / R_{\theta JA}$. All numbers apply for packages soldered directly onto a PCB.

7.2 ESD Ratings

The following ESD information is provided for handling of ESD-sensitive devices in an ESD protected area only.

			VALUE	UNIT
$V_{(ESD)}$	Electrostatic discharge	Human-body model (HBM)	± 3000	V
		Charged-device model (CDM)	± 1000	
		Machine Model (MM)	± 400	



ESD SENSITIVITY CAUTION

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

7.3 Recommended Operating Conditions

Over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
Supply voltage, $V_S= (V+) - (V-)$	Single-supply	1.4		5.5	V
	Dual-supply	± 0.7		± 2.75	

7.4 ELECTRICAL CHARACTERISTICS

(At $T_A = +25^\circ\text{C}$, $V_S = 1.4\text{V}$ to 5.5V , $V_{CM} = V_S/2$, $C_L = 15\text{pF}$, unless otherwise noted.)⁽¹⁾

PARAMETER		CONDITIONS		RS8905/RS8905B			
				MIN ⁽²⁾	TYP ⁽³⁾	MAX ⁽²⁾	UNITS
POWER SUPPLY							
V_S	Operating Voltage Range			1.4		5.5	V
I_Q /AMP	Quiescent Current Per channel				400	1500	nA
PSRR	Power-Supply Rejection Ratio	$V_S = 1.4\text{V}$ to 5.5V , $V_{CM} = (V) + 0.5\text{V}$			70		dB
INPUT							
V_{OS}	Input Offset Voltage	$V_{CM} = V_S/2$	$V_S = 1.4\text{V}$	-6	1	6	mV
			$V_S = 5.0\text{V}$	-3	1	3	
$\Delta V_{OS}/\Delta T$	Input Offset Voltage Drift	$V_{CM} = V_S/2$, $-40^\circ\text{C} \leq T_A \leq 125^\circ\text{C}$			2		$\mu\text{V}/^\circ\text{C}$
I_B	Input Bias Current ⁽⁴⁾⁽⁵⁾				1	10	pA
V_{CM}	Common-Mode Voltage Range	$T_A = -40^\circ\text{C}$ to 125°C		(V-)-0.1		(V+)+0.1	V
CMRR	Common-Mode Rejection Ratio	$V_S = 5.5\text{V}$, $V_{CM} = -0.1$ to 5.6V			70		dB
OUTPUT							
V_{OH}	Output Swing From Upper Rail	$V_S = 1.4\text{V}$, $I_O = 0.1\text{mA}$			70	75	mV
		$V_S = 5.0\text{V}$, $I_O = 2.5\text{mA}$			140	170	mV
V_{OL}	Output Swing From Lower Rail	$V_S = 1.4\text{V}$, $I_O = -0.1\text{mA}$			35	40	mV
		$V_S = 5.0\text{V}$, $I_O = -2.5\text{mA}$			85	115	mV
I_{SC}	Short Circuit Sink Current	$V_S = 5.0\text{V}$			42		mA
	Short Circuit Source Current	$V_S = 5.0\text{V}$			38		mA
SWITCHING							
T_{PHL}	Propagation Delay H To L ⁽⁶⁾	$V_S = 5.0\text{V}$, Overdrive = 10 mV			13	20	μs
		$V_S = 5.0\text{V}$, Overdrive = 100 mV			9	14	
		$V_S = 2.5\text{V}$, Overdrive = 10 mV			12	18	
		$V_S = 2.5\text{V}$, Overdrive = 100 mV			8	12	
		$V_S = 1.4\text{V}$, Overdrive = 10 mV			13	20	
		$V_S = 1.4\text{V}$, Overdrive = 100 mV			9	14	
T_{PLH}	Propagation Delay L To H ⁽⁶⁾	$V_S = 5.0\text{V}$, Overdrive = 10 mV			30	50	
		$V_S = 5.0\text{V}$, Overdrive = 100 mV			21	38	
		$V_S = 2.5\text{V}$, Overdrive = 10 mV			24	45	
		$V_S = 2.5\text{V}$, Overdrive = 100 mV			15	30	
		$V_S = 1.4\text{V}$, Overdrive = 10 mV			25	50	
		$V_S = 1.4\text{V}$, Overdrive = 100 mV			15	30	
T_R	Rise Time	Overdrive = 100 mV			240		ns
T_F	Fall Time	Overdrive = 100 mV			260		ns

NOTE:

- (1) Electrical table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device.
- (2) Limits are 100% production tested at 25°C . Limits over the operating temperature range are ensured through correlations using statistical quality control (SQC) method.
- (3) Typical values represent the most likely parametric norm as determined at the time of characterization. Actual typical values may vary over time and will also depend on the application and configuration.
- (4) This parameter is ensured by design and/or characterization and is not tested in production.
- (5) Positive current corresponds to current flowing into the device.
- (6) High-to-low and low-to-high refers to the transition at the input.

7.5 TYPICAL CHARACTERISTICS

NOTE: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only.

At $T_A = +25^\circ\text{C}$, $V_S=5\text{V}$, $V_{CM} = V_S/2$, $C_L=15\text{pF}$ unless otherwise noted.

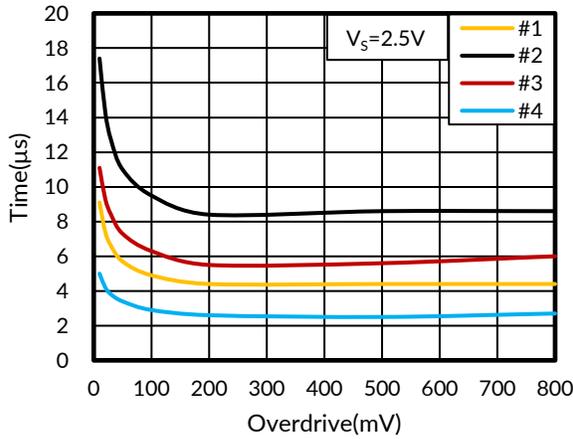


Figure 1. Response Time vs Input Overdrives Negative Transition

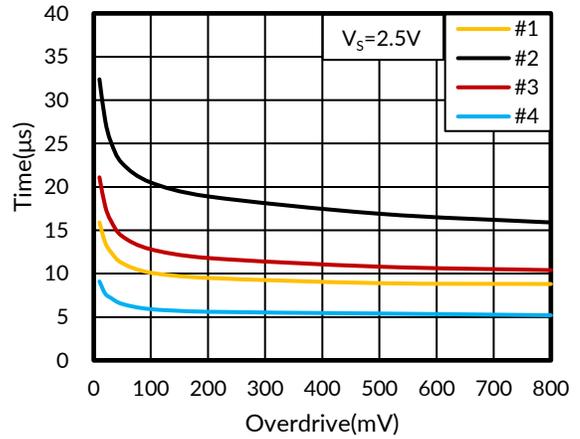


Figure 2. Response Time vs Input Overdrives Positive Transition

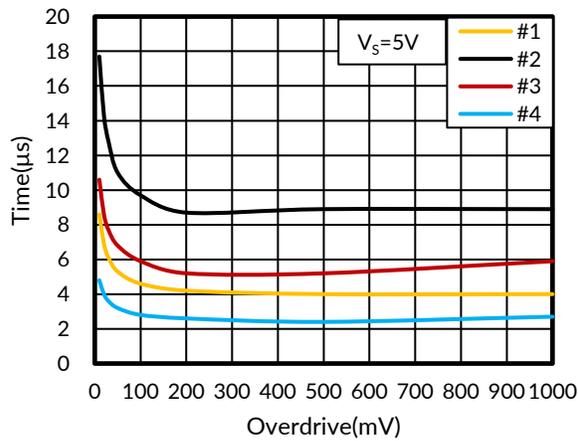


Figure 3. Response Time vs Input Overdrives Negative Transition

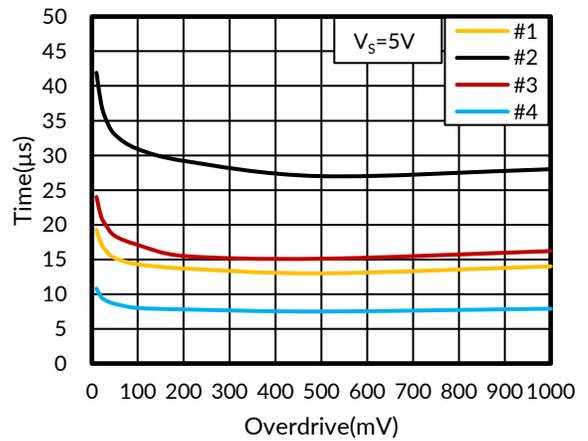


Figure 4. Response Time vs Input Overdrives Positive Transition

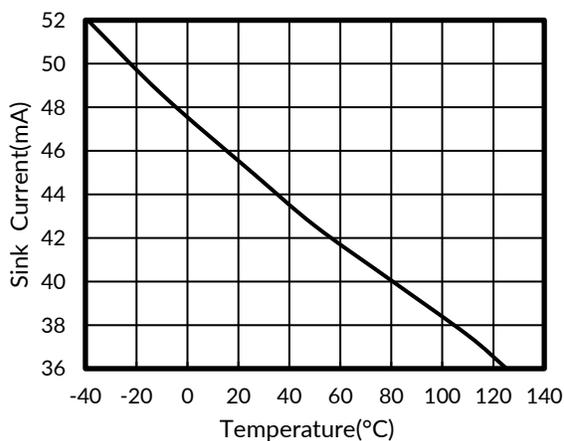


Figure 5. SINK CURRENT vs TEMPERATURE

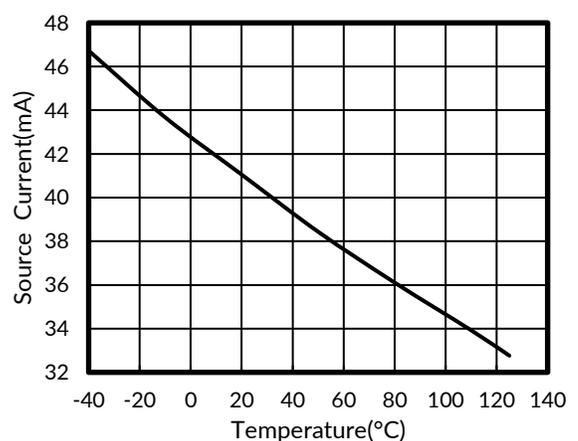


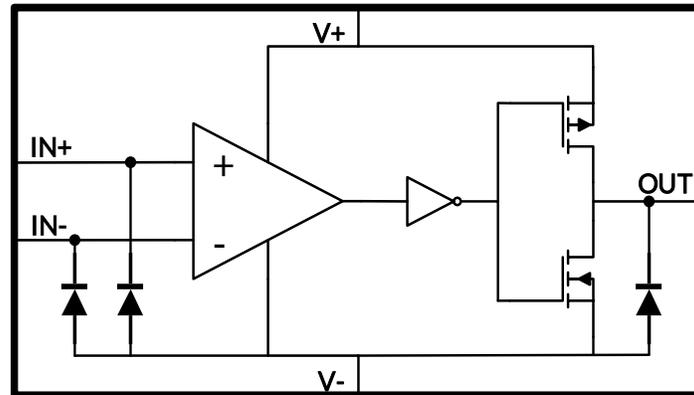
Figure 6. SOURCE CURRENT vs TEMPERATURE

8 DETAILED DESCRIPTION

8.1 Overview

The RS8905 devices are double-channel, Nano-power comparators with a push-pull output stage. Operating from 1.4V to 5.5V and consuming only 400nA per channel. The push-pull output of the RS8905 supports rail-to-rail output swing and interfaces with TTL/CMOS logic.

8.2 Functional Block Diagram



8.3 Feature Description

The RS8905 devices are Nano-Power comparators that can operate at low voltages. The RS8905 feature a rail-to-rail input stage capable of operating up to 100 mV beyond the VCC power supply rail.

8.4 Input Stage

The RS8905 has rail-to-rail input common-mode voltage range. It can operate at any differential input voltage within this limit as long as the differential voltage is greater than zero. A differential input of zero volts may result in oscillation.

The differential input stage of the comparator is a pair of PMOS and NMOS transistors, therefore, no current flows into the device. The input bias current measured is the leakage current in the MOS transistors and input protection diodes. This low bias current allows the comparator to interface with a variety of circuitry and devices with minimal concern about matching the input resistances.

8.5 Output Stage

The RS8905 has a MOS push-pull rail-to-rail output stage. The push-pull transistor configuration of the output keeps the total system power consumption to a minimum. The only current consumed by the RS8905 is the less than 1 μ A supply current and the current going directly into the load. No power is wasted through the pullup resistor when the output is low. The output stage is specifically designed with dead time between the time when one transistor is turned off and the other is turned on (break-before-make) to minimize shoot through currents. The internal logic controls the break-before-make timing of the output transistors. The break-before-make delay varies with temperature and power condition.

8.6 Output Current

Even though the RS8905 uses less than 1 μ A supply current, the outputs are able to drive very large currents. The RS8905 can source up to 38mA and can sink up to 42mA, when operated at 5V supply. This large current handling capability allows driving heavy loads directly.

9 APPLICATION INFORMATION

The RS8905 is an ultra-low-power comparator with a typical power supply current of 400nA per channel. It has the best-in class power supply current versus propagation delay. The propagation delay is as low as 9μs with 100mV overdrive at 1.4V supply.

Typical Applications

9.1 Square Wave Generator

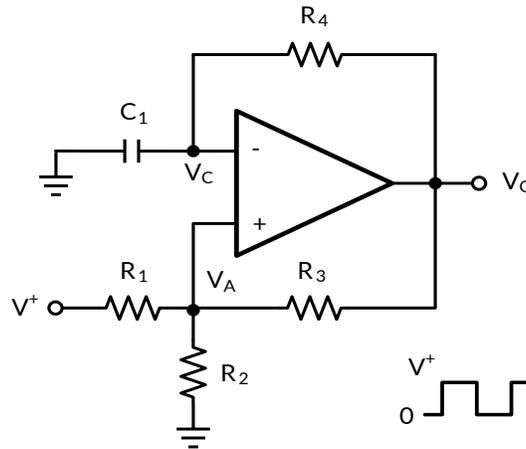


Figure 7. Square Wave Generator Schematic

9.2 Design Requirements

A typical application for a comparator is as a square wave oscillator. The circuit in Figure 7 generates a square wave whose period is set by the RC time constant of the capacitor C1 and resistor R4. The maximum frequency is limited by the large signal propagation delay of the comparator and by the capacitive loading at the output, which limits the output slew rate.

9.3 Detailed Design Procedure

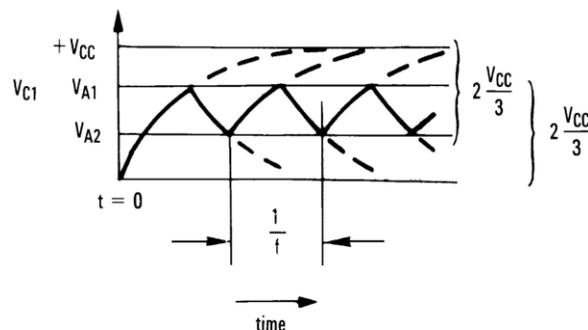


Figure 8. Square Wave Oscillator

Typical Applications(continued)

Consider the output of Figure 8 to be high to analyze the circuit. That implies that the inverted input (V_C) is lower than the noninverting input (V_A). This causes the C_1 to be charged through R_4 , and the voltage V_C increases until it is equal to the noninverting input. The value of V_A at this point is in Equation 1.

$$V_{A1} = \frac{V_{CC} \times R_2}{R_2 + R_1 \parallel R_3} \dots\dots\dots (1)$$

If $R_1 = R_2 = R_3$ then $V_{A1} = 2 V_{CC}/3$

At this point the comparator switches pulling down the output to the negative rail. The value of V_A at this point, as shown in Equation 2:

$$V_{A2} = \frac{V_{CC}(R_2 \parallel R_3)}{R_1 + (R_2 \parallel R_3)} \dots\dots\dots (2)$$

If $R_1 = R_2 = R_3$ then $V_{A2} = V_{CC}/3$ The capacitor C_1 now discharges through R_4 , and the voltage V_C decreases until it is equal to V_{A2} , at which point the comparator switches again, bringing it back to the initial stage. The time period is equal to twice the time it takes to discharge C_1 from $2 V_{CC}/3$ to $V_{CC}/3$, which is given by $R_4 C_1 \times \ln 2$. Hence the formula for the frequency is given by Equation 3:

$$F = 1 / (2 \times R_4 \times C_1 \times \ln 2) \dots\dots\dots (3)$$

9.4 Application Curves

Figure 9 shows the simulated results of an oscillator using the following values:

1. $R_1 = R_2 = R_3 = R_4 = 100 \text{ k}\Omega$
2. $C_1 = 100 \text{ pF}$, $C_L = 20 \text{ pF}$
3. $V_+ = 5 \text{ V}$, $V_- = \text{GND}$
4. C_{STRAY} (not shown) from V_a to $\text{GND} = 10 \text{ pF}$

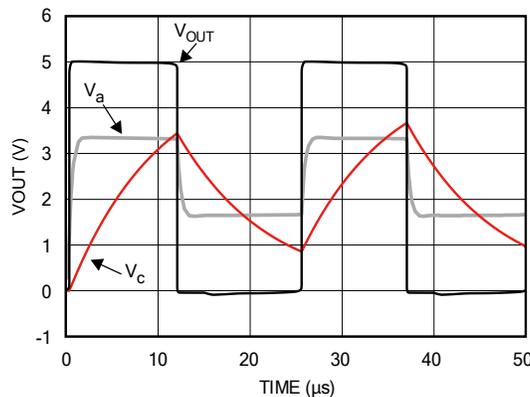
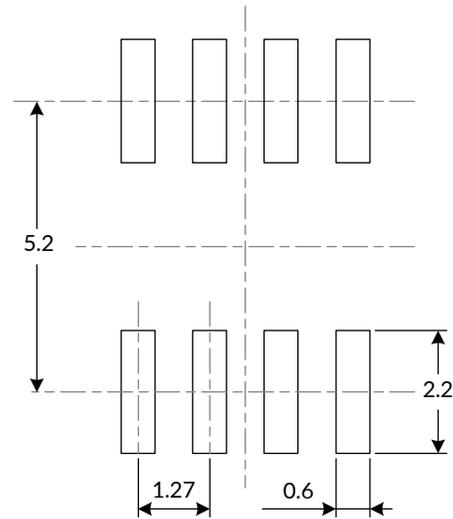
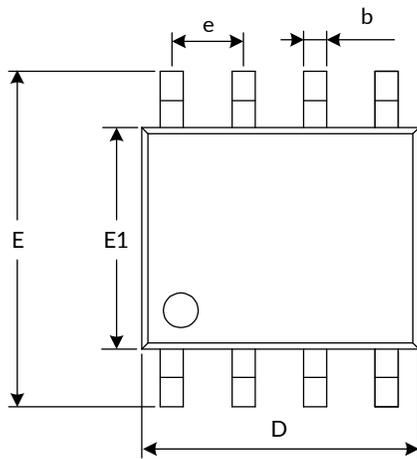
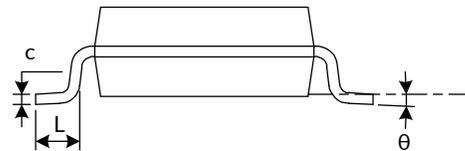
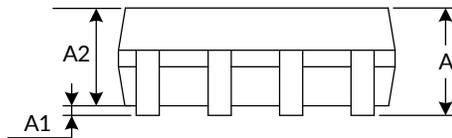


Figure 9. Square Wave Oscillator Output Waveform

10 PACKAGE OUTLINE DIMENSIONS

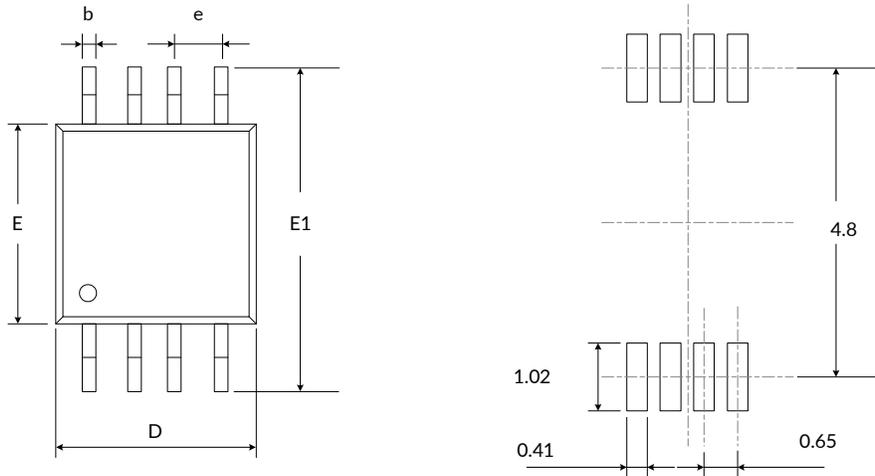
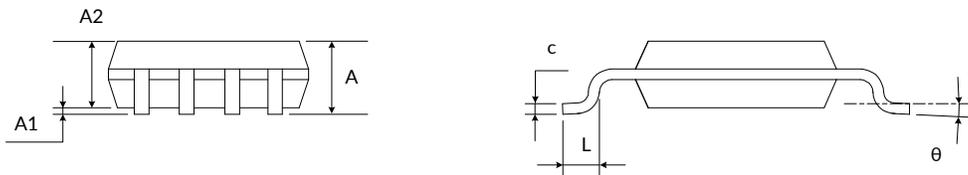
SOP8⁽³⁾


RECOMMENDED LAND PATTERN (Unit: mm)


Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A ⁽¹⁾	1.350	1.750	0.053	0.069
A1	0.100	0.250	0.004	0.010
A2	1.350	1.550	0.053	0.061
b	0.330	0.510	0.013	0.020
c	0.170	0.250	0.007	0.010
D ⁽¹⁾	4.800	5.000	0.189	0.197
e	1.270 (BSC) ⁽²⁾		0.050 (BSC) ⁽²⁾	
E	5.800	6.200	0.228	0.244
E1 ⁽¹⁾	3.800	4.000	0.150	0.157
L	0.400	1.270	0.016	0.050
θ	0°	8°	0°	8°

NOTE:

1. Plastic or metal protrusions of 0.15mm maximum per side are not included.
2. BSC (Basic Spacing between Centers), "Basic" spacing is nominal.
3. This drawing is subject to change without notice.

MSOP8⁽³⁾

RECOMMENDED LAND PATTERN (Unit: mm)


Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A ⁽¹⁾	0.820	1.100	0.032	0.043
A1	0.020	0.150	0.001	0.006
A2	0.750	0.950	0.030	0.037
b	0.250	0.380	0.010	0.015
c	0.090	0.230	0.004	0.009
D ⁽¹⁾	2.900	3.100	0.114	0.122
e	0.650 (BSC) ⁽²⁾		0.026 (BSC) ⁽²⁾	
E ⁽¹⁾	2.900	3.100	0.114	0.122
E1	4.750	5.050	0.187	0.199
L	0.400	0.800	0.016	0.031
θ	0°	6°	0°	6°

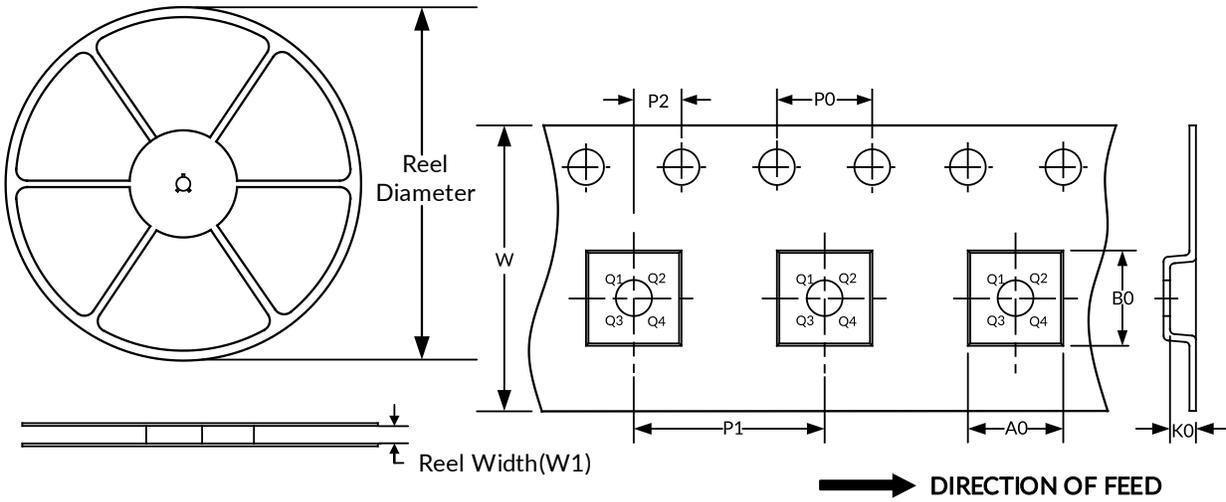
NOTE:

1. Plastic or metal protrusions of 0.15mm maximum per side are not included.
2. BSC (Basic Spacing between Centers), "Basic" spacing is nominal.
3. This drawing is subject to change without notice.

11 TAPE AND REEL INFORMATION

REEL DIMENSIONS

TAPE DIMENSION



NOTE: The picture is only for reference. Please make the object as the standard.

KEY PARAMETER LIST OF TAPE AND REEL

Package Type	Reel Diameter	Reel Width(mm)	A0 (mm)	B0 (mm)	K0 (mm)	P0 (mm)	P1 (mm)	P2 (mm)	W (mm)	Pin1 Quadrant
SOP8	13"	12.4	6.40	5.40	2.10	4.0	8.0	2.0	12.0	Q1
MSOP8	13"	12.4	5.20	3.30	1.50	4.0	8.0	2.0	12.0	Q1

NOTE:

1. All dimensions are nominal.
2. Plastic or metal protrusions of 0.15mm maximum per side are not included.

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